

ETH zürich



Empa

Materials Science and Technology

Acoustics II

Audio effects

Reto Pieren

2024

Audio effects

[Zölzer, Udo. et al. 2002. DAFX – Digital Audio Effects. John Wiley & Sons.

Weinzierl, Stefan 2009. Handbuch der Audiotechnik, Springer.

Pohlmann, K.C. 2005. Principles of Digital Audio. McGraw-Hill.

Smith III, Julius O. 2010. Physical Audio Signal Processing: for Virtual Musical Instruments and Digital Audio Effects. W3K Publishing.]

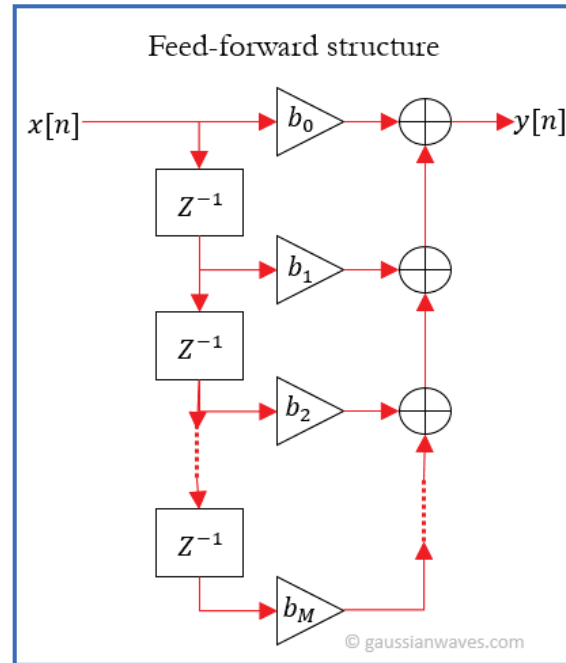
Audio effects: Introduction

- Audio signal processing
- Today mostly digital audio signal processing (PC or DSP)
- Alteration of the original sound for:
 - adjustment of a transmission effect (e.g. frequency response)
 - reduction of the audibility of unwanted signal components
 - mimiking a source behavior or location
 - ...
 - creation of new sounds



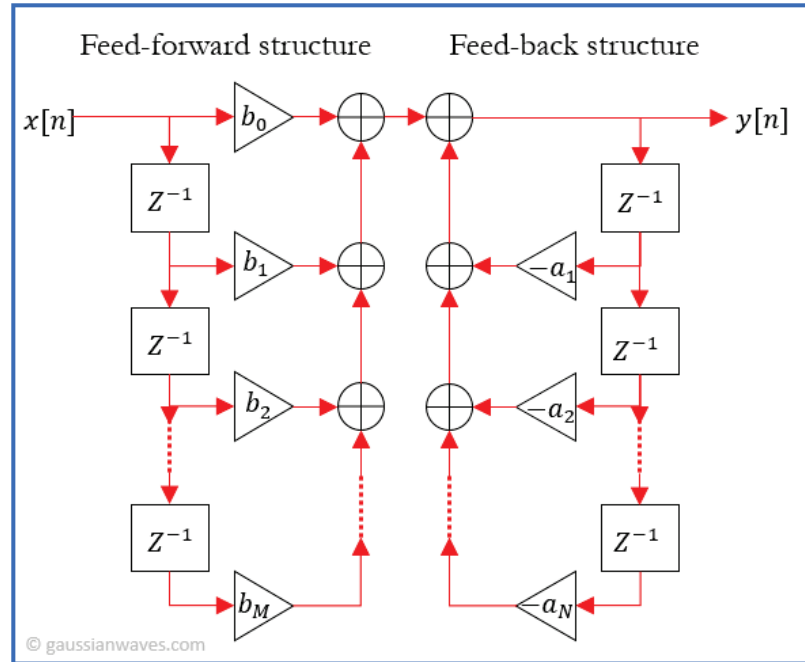
Digital filters: Finite Impulse Response (FIR)

- Finite IR length
- Only forward processing
- Always stable
- Computationally expensive
- Simpler to design
- Order = number of delays = IR length - 1



Digital filters: Infinite Impulse Response (IIR)

- Contains feedback loops
- Generally IR does not drop to 0
- Compact
- Efficient
- May be unstable
- Order = min. number of needed delay elements



Equalizer

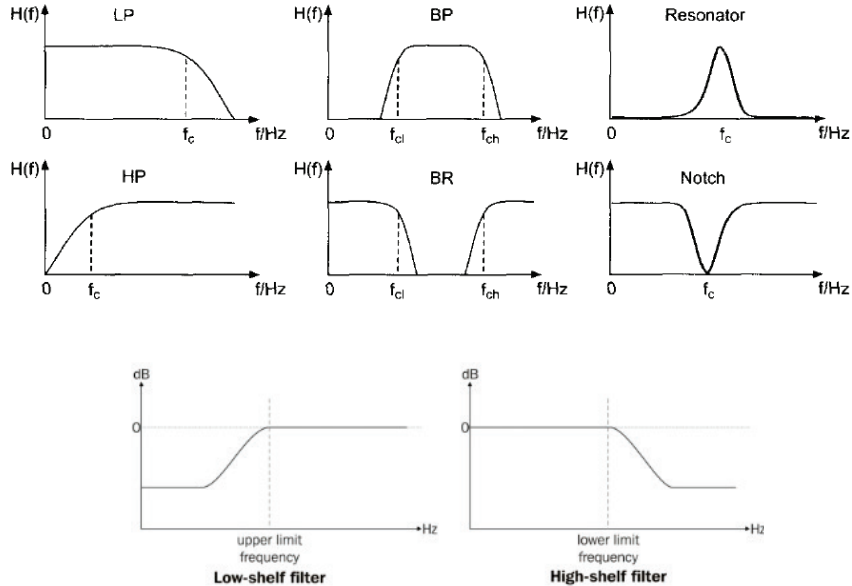
Equalizer (EQ)

- Function:
 - manipulation of the amplitude response of a sound or a transmission system
- Applications:
 - Tone control
 - flatten a non-ideal frequency response of a loudspeaker
 - loudness filtering
 - equalization of headphones
 - attenuation of the low frequency end in PA systems for speech
- LTI system = full characterization by frequency/impulse response

[Välimäki, Vesa & Reiss, J.D. 2016. All About Audio Equalization: Solutions and Frontiers. *Applied Sciences* 2016(6).]



Equalizer (EQ): Elementary filter types



- Low-pass and High-pass (LP, HP)
- Bandpass and Bandstop (BP, BS)
- Peaking/Bell and Notch
- Low shelf and High shelf
- All-pass = phase equalizer

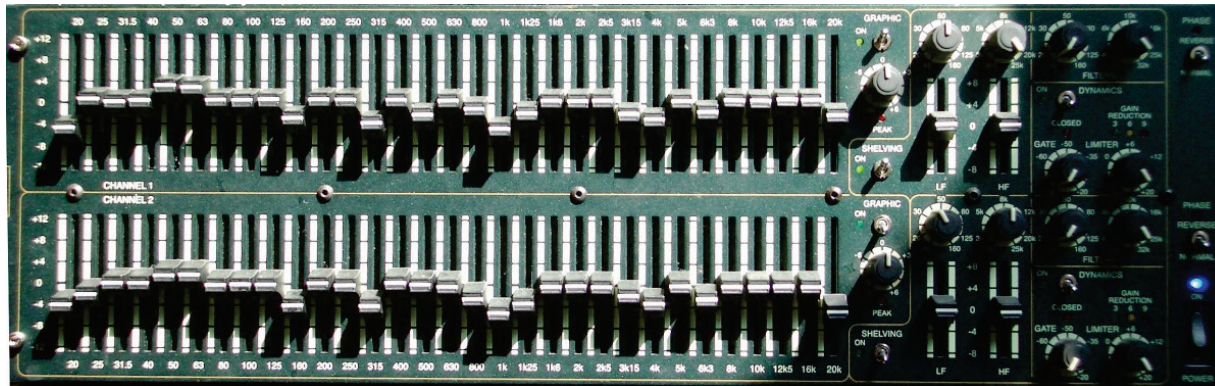
$$|H(j\omega)| = 1$$

First order all-pass:

$$H(j\omega) = \frac{j\omega - \omega_0}{j\omega + \omega_0}$$

Equalizer (EQ): Graphical equalizer

- Parallel bandpass or cascaded bell filters of constant relative bandwidth
 - 1/3-octave filters (typ. 31 bands), $Q=4$
 - octave filters (typ. 10 bands), $Q=1.4$
- Adjustable amplification/attenuation in each band (typ. ± 12 dB)
- Graphical representation of the amplitude response









Equalizer (EQ): Adjustments in 1/3 octave bands

- Listening test: pink noise, amplification of the 800 Hz third octave band with normalization of total level
- → answers: *clearly -*, *just -*, *not audible*

1	A		B	
2	A		B	
3	A		B	
4	A		B	

Equalizer (EQ): Adjustments in 1/3 octave bands

- Listening test: pink noise, amplification of the 800 Hz third octave band with normalization of total level
- → answers: *clearly -*, *just -*, *not audible*

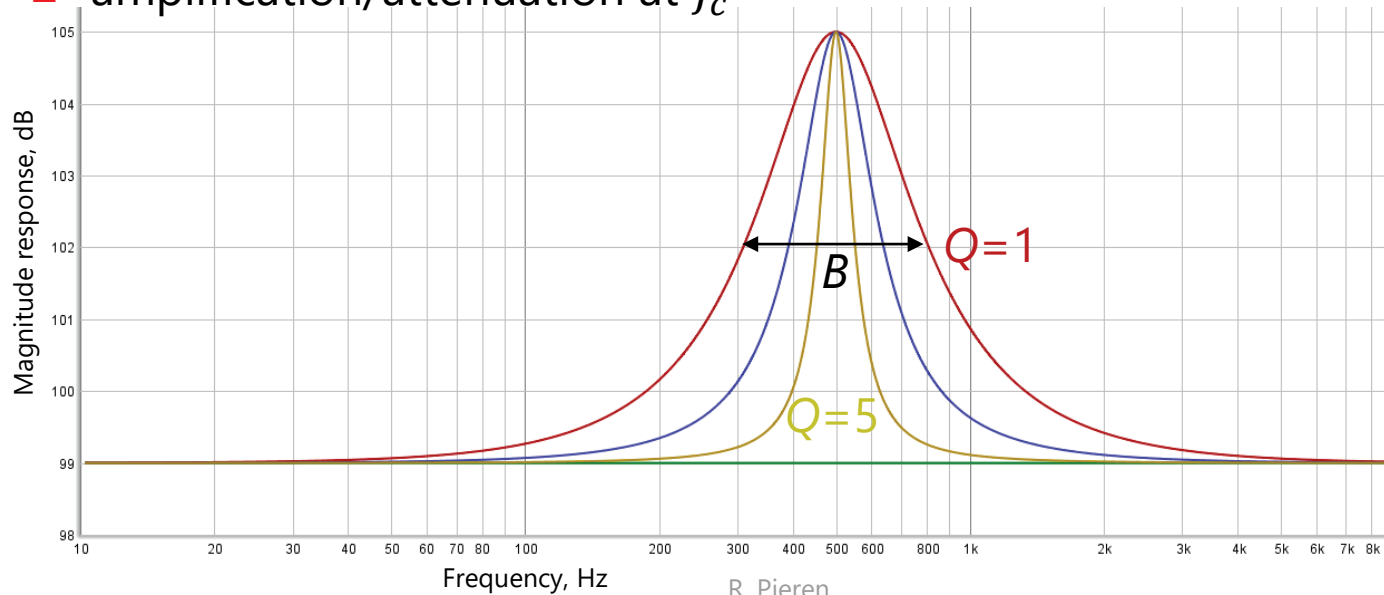
1	A: flat 	B: +9 dB 
2	A: flat 	B: flat 
3	A: flat 	B: +3 dB 
4	A: flat 	B: +6dB 

Equalizer (EQ): Parametric equalizer

- Cascade of bell filters with variable

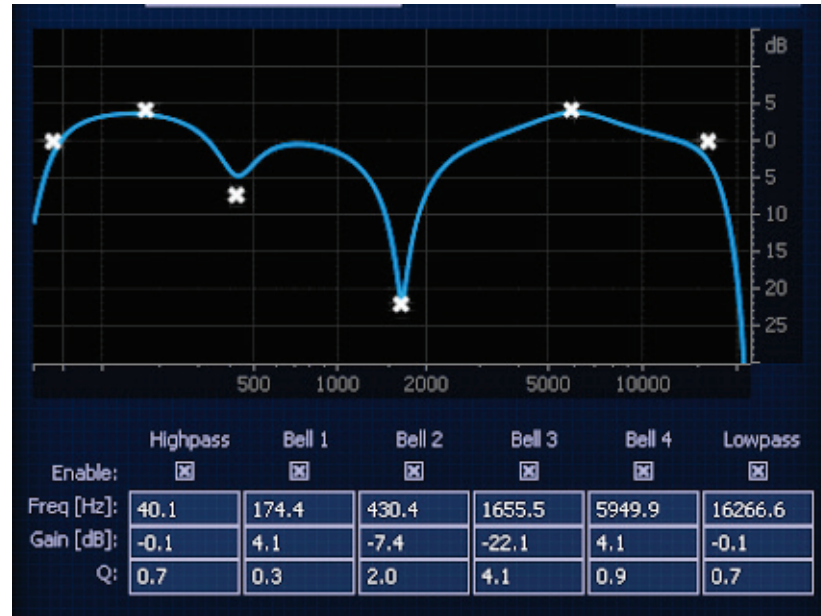
- center frequency f_c
- bandwidth B , or $Q = f_c/B$
- amplification/attenuation at f_c

$$Q = \frac{2^{n/2}}{2^n - 1}, n\text{-octave bandwidth}$$



Equalizer (EQ): Parametric equalizer

- Example of software implementation
 - Bells + highpass and lowpass

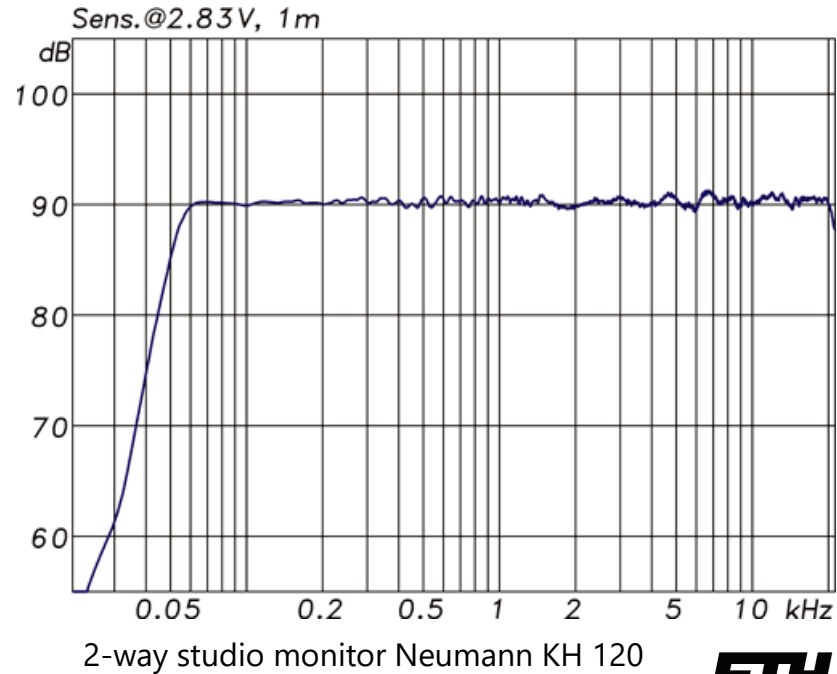


Equalizer (EQ): Parametric equalizer

- Universal applicability
- Not that intuitive to reproduce a predefined amplitude response
- Can be easily made time-variable if realized by IIR filters

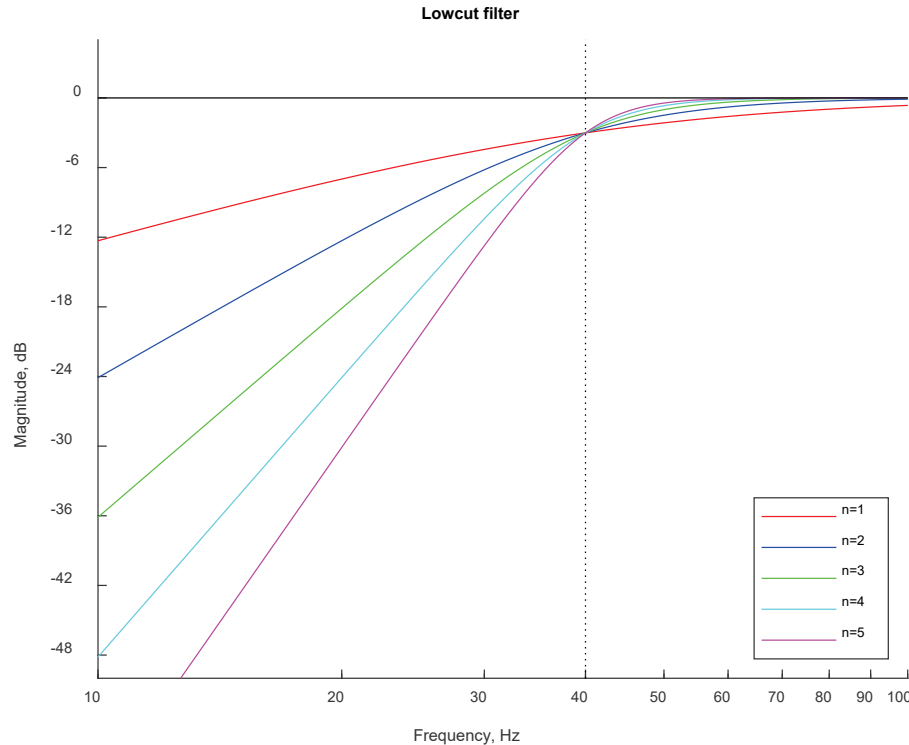
Equalizer (EQ): Example Low cut

- Application
 - Assumption: Playback system limited by lower cutoff frequency of 50 Hz
 - Signal components below
 - unnecessary burden for amplifier
 - limiting signal dynamic range
- Low cut filter



Equalizer (EQ): Example Low cut

- Often used type: Butterworth high-pass filter



- Maximally flat in passband

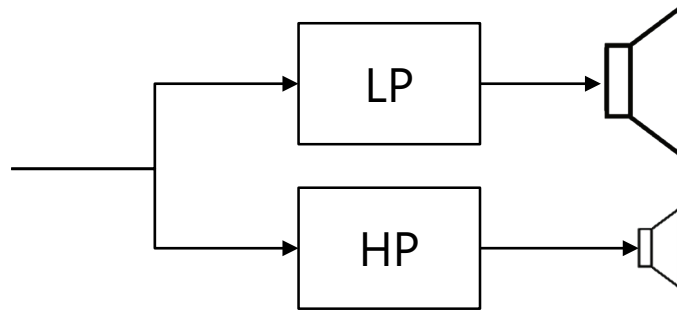
- Slope = $6n$ dB/octave

n : filter order

$$f_c = 40 \text{ Hz}$$

Equalizer (EQ): Example Crossover

- Application:
 - Subwoofers for low frequency content
 - Crossover filtering for subwoofers and satellites at ~ 100 Hz



- First-order filtering (Butterworth=Bessel):

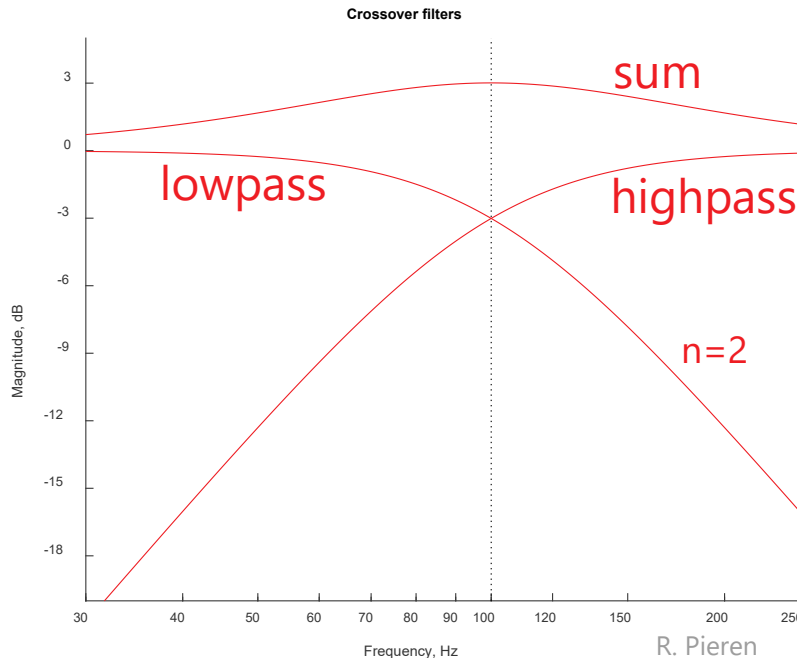
$$H_{LP} = \frac{\omega_0}{s+\omega_0}; H_{HP} = \frac{s}{s+\omega_0}; H_{tot} = H_{LP} + H_{HP} = \frac{s+\omega_0}{s+\omega_0} = 1$$

→ sums up to 1 (ideal solution)

(90° phase shift between outputs)

Equalizer (EQ): Example Crossover

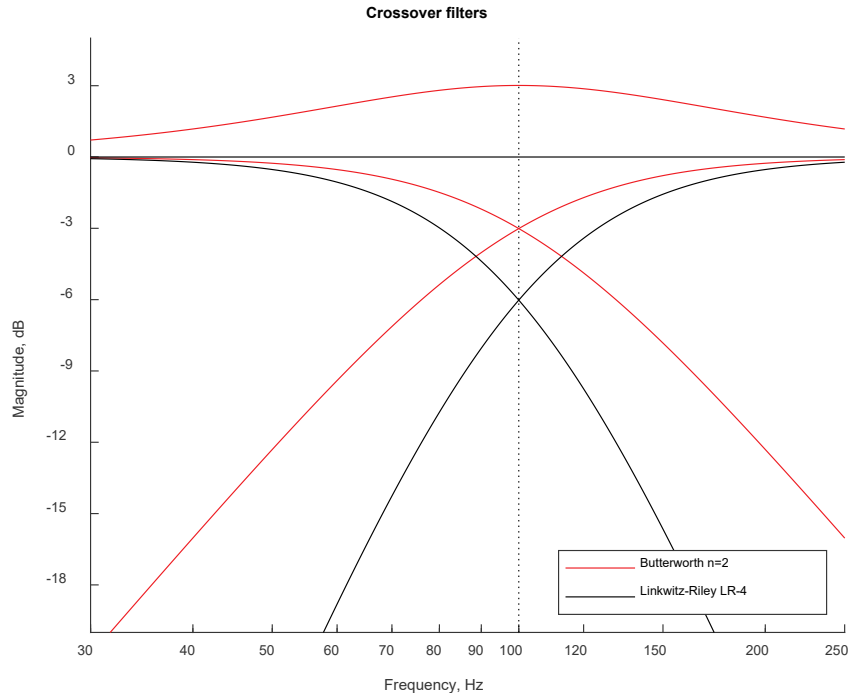
- Smaller transition region required \rightarrow steeper filters \rightarrow higher order filters
- Butterworth filters?



- Sum at crossover not 0 dB
- phase shifts between outputs (multiple of 180°)

Equalizer (EQ): Example Crossover

- Linkwitz–Riley (LR)-filter = 2 cascaded Butterworth



- Sum at crossover = 0dB
- Outputs in phase
- 4th order LR crossover = cascade of two 2nd order Butterworth → 24 dB/oct

[Linkwitz, S.H. 1976. Active crossover networks for noncoincident drivers. *Journal of the Audio Engineering Society* 24(1).

<https://www.ranecommercial.com/legacy/note160.html>]

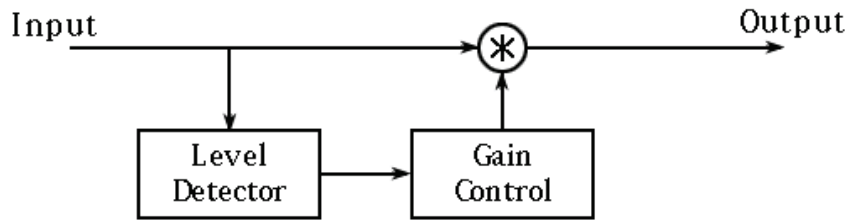
Dynamics

Compressor

Compressor/limiter

- Function:
 - reduction of signal dynamics → non-linear operation
- Applications for compressors:
 - increase of loudness with given maximal signal amplitude
 - music production
 - commercials
 - radio stations (listeners are often in noisy environment)
 - homogenization of unstable signal
 - artistic feature (e.g. audibility of breathing in a voice)
- Applications for limiters: avoid clipping in digital recordings

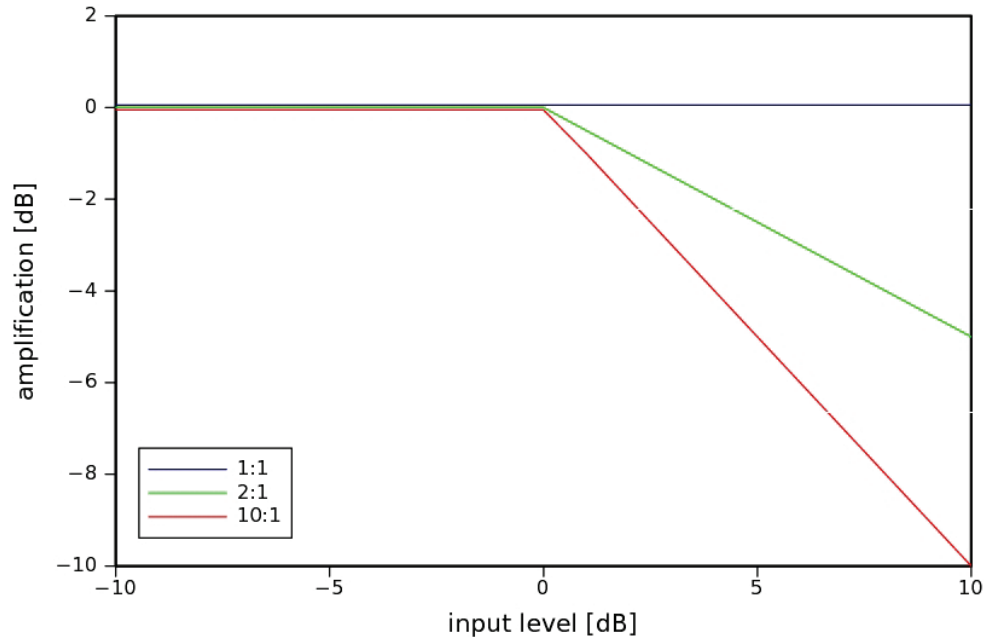
Compressor/limiter: Principle of operation



- Analog: Voltage Controlled Amplifier
- Steering voltage: *smoothed* input signal
- Reduced amplification above an adjustable threshold

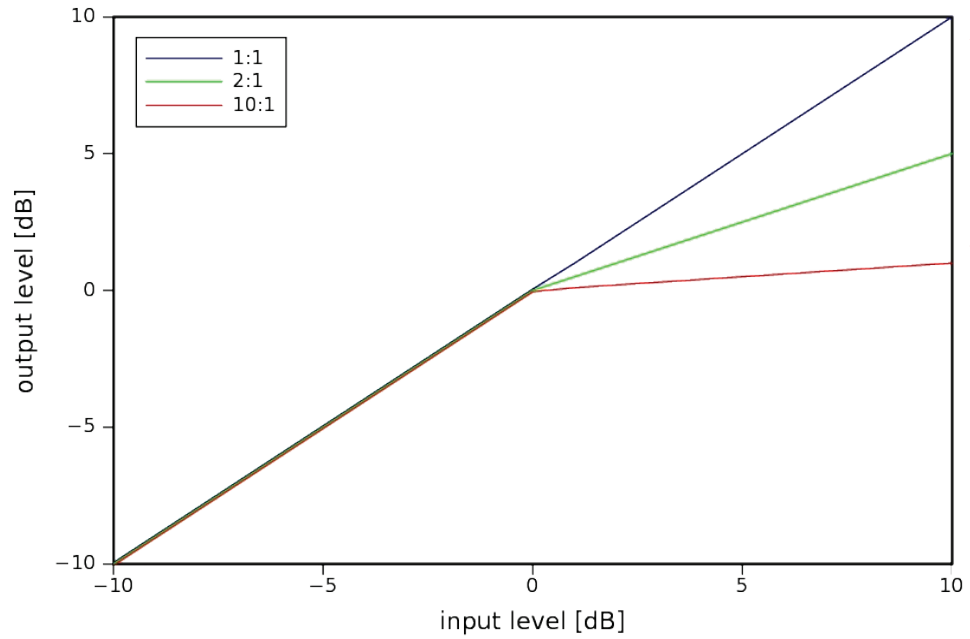
Compressor/limiter: Characteristics

- Parameters: Level *threshold* (here = 0dB) and *ratio*
- Input level → Amplification



Compressor/limiter: Characteristics

- Parameters: Level *threshold* (here = 0 dB) and *ratio*
- Input level → Output level



10:1
9 dB attenuation
at 10 dB above
threshold

■ Limiter: $\infty:1$

Compressor/limiter: Time constants

- Amplification is steered by the level of the audio signal
- Determination of level needs an averaging process
- Different averaging types (*RMS*, *peak*)
- Typical use of two time constants:
 - *Attack* time: for signal increase (positive slopes)
 - *Release* time: for signal decrease (negative slopes)

Compressor/limiter: Time constants

- Choice of constants is critical
 - ! Attack too small → distortion of transients
 - ! Release too small → pumping: modulation of signal by dominant components
 - ! Attack/release too large → signal dynamics is not properly reduced due to delayed reaction
- Typical values (highly dependent on audio material)
 - Attack: order of milliseconds
 - Release: 10 – 3000 ms

Compressor/limiter: Example

■ Original sample: 

■ Compression starting 25 dB below full-scale
→ approx. 10 dB higher signal power at identical peak value:

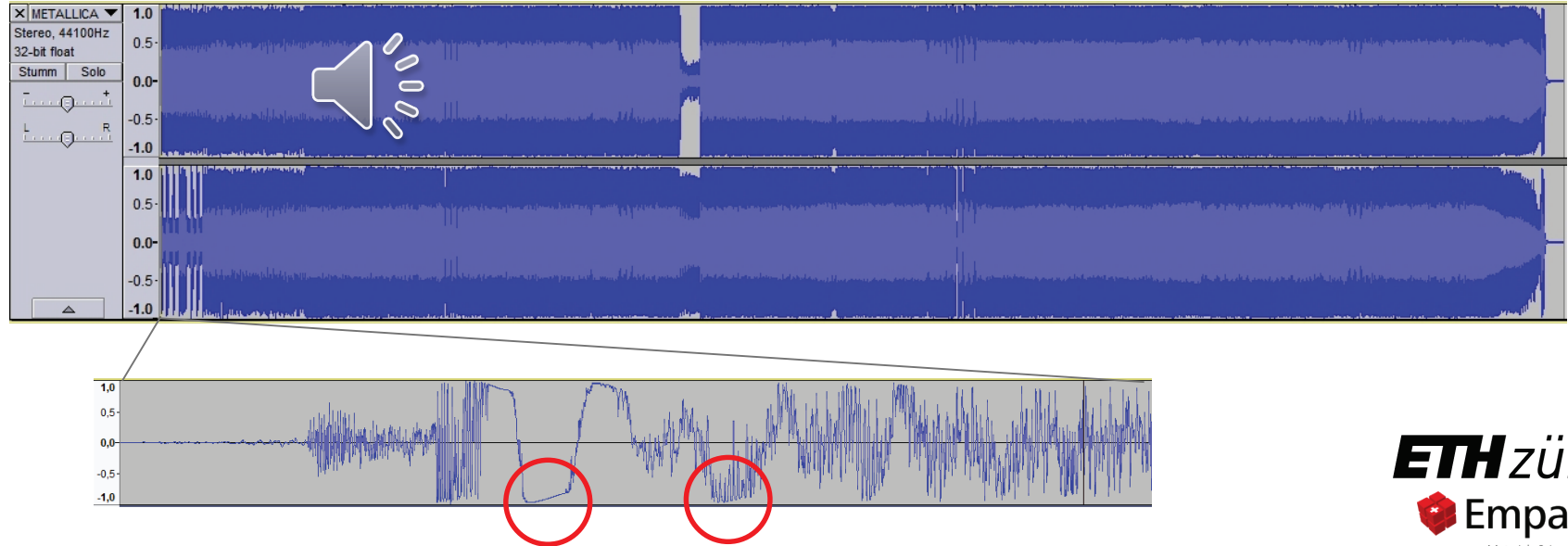


Compressor/limiter: Finalizer

- *Loudness war*
 - producers push perceived loudness to maximum (radio, TV, music)
 - increasing concern regarding audio quality
 - loss in dynamic range
 - unofficial dynamic range data base:
<http://www.dr.loudness-war.info/>

Compressor/limiter: Finalizer

- Loudness war example:
 - CD version of *Metallica: Death Magnetic: My Acopalypse*, 2008



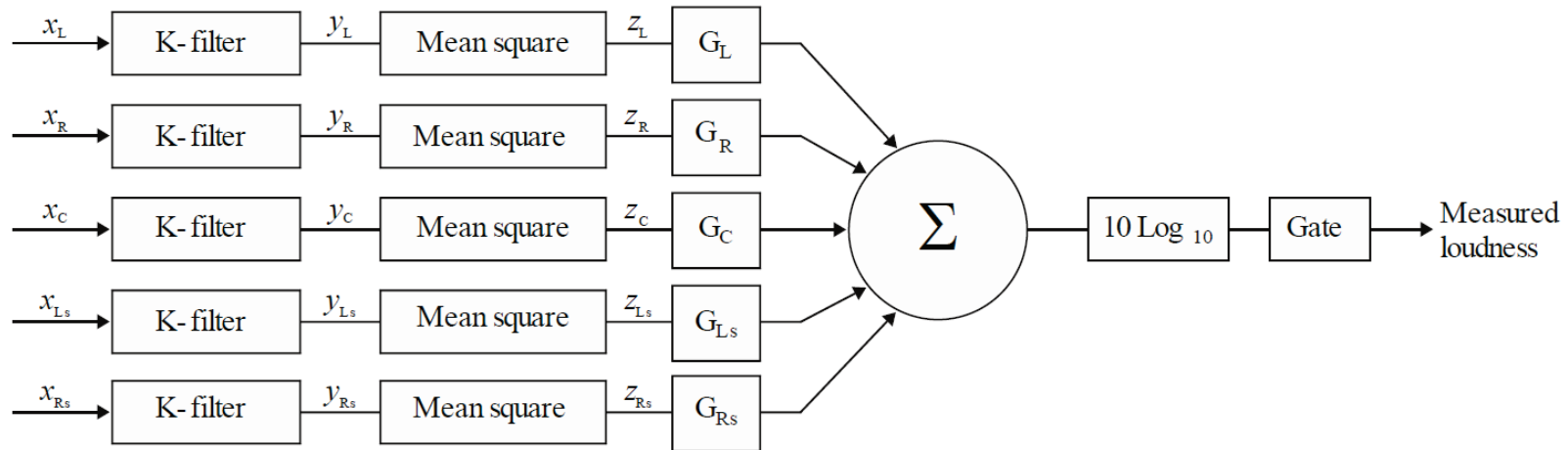
Compressor/limiter: Loudness normalization

- End of the loudness war? → Loudness normalization
 - Evaluate loudness level of each track → adjust playback gain for each track to unified loudness level
 - Applied in today's online portals, e.g. Spotify or YouTube
 - Loudness L_{KG} measured in LKFS/LUFS (Loudness Units relative to Full Scale) (equal to dB)

[ITU 2015. Recommendation ITU-R BS.1770-4 - Algorithms to measure audio programme loudness and true-peak audio level, International Telecommunication Union (ITU), Geneva, Switzerland.]

Compressor/limiter: Loudness normalization

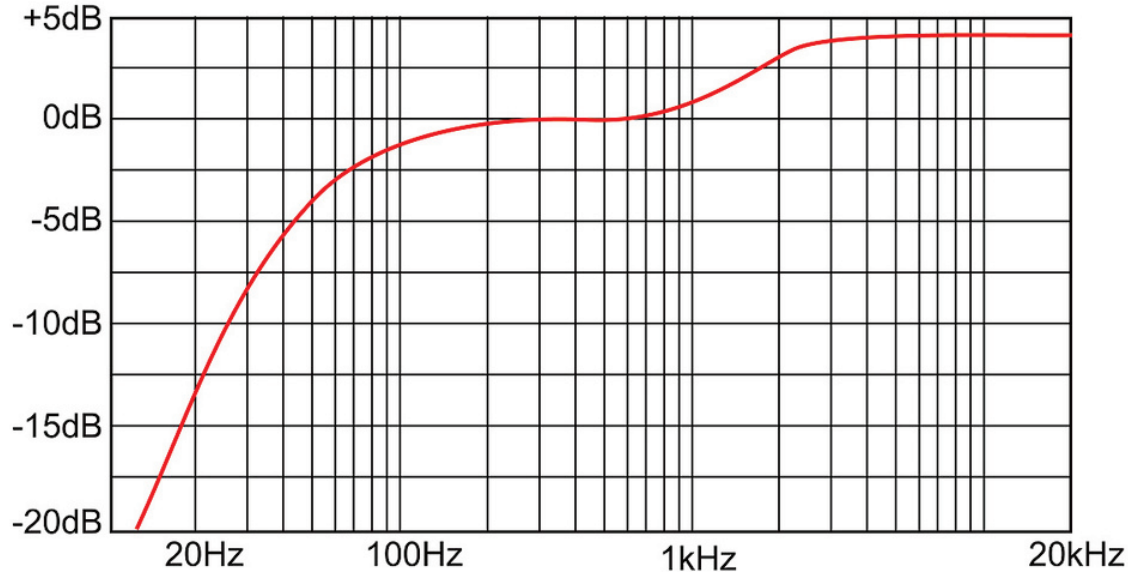
Simplified block diagram of multichannel loudness algorithm



[ITU 2015. Recommendation ITU-R BS.1770-4 -
Algorithms to measure audio programme loudness and true-peak audio level,
International Telecommunication Union (ITU), Geneva, Switzerland.]

Compressor/limiter: Loudness normalization

K-Weighting Filter Response



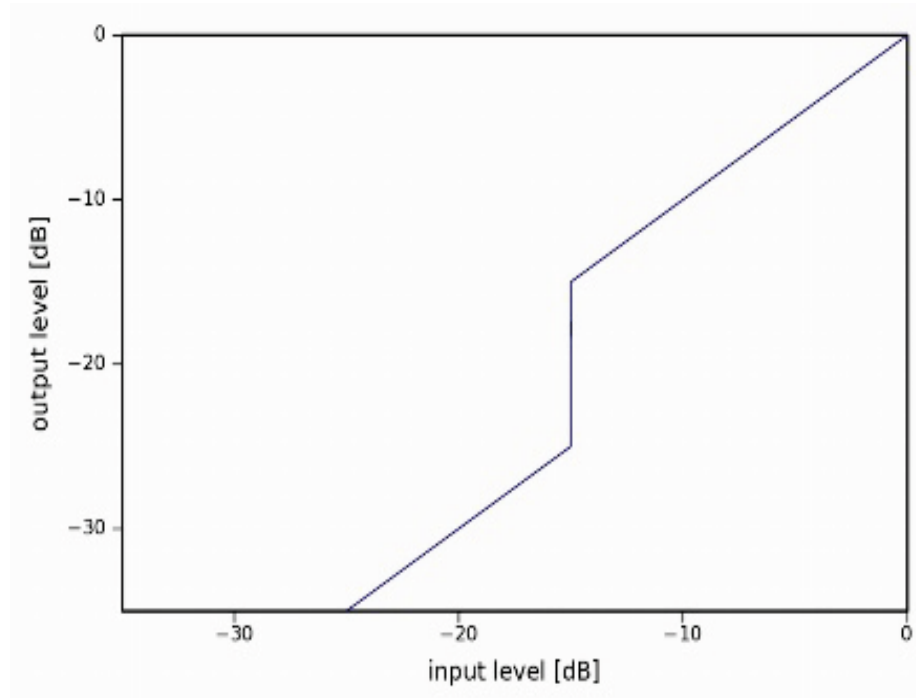
Noise gate

Noise gate

- Function
 - Attenuation of low level signals
- Application
 - Suppression of noise during signal pauses

Noise gate

- Parameters: Level threshold + step



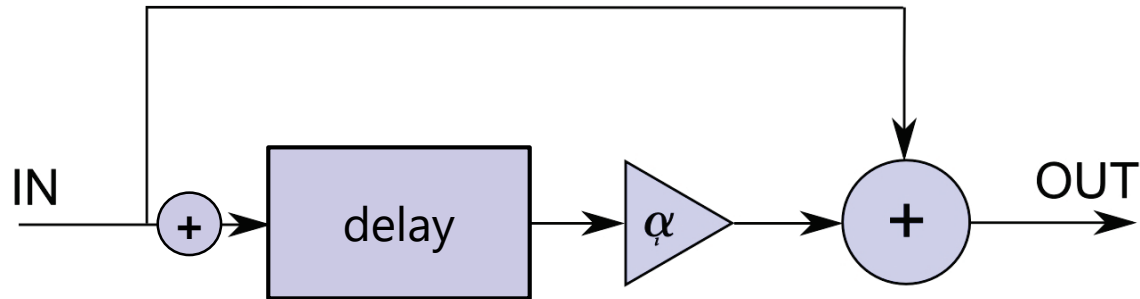
Advanced dynamics and EQ effects

- Various more advanced (combined) effects
 - Side-chain compressor
 - Multiband compressor
 - Dynamic EQ
 - Exciter
 - Wah-wah
 - ...

Delays

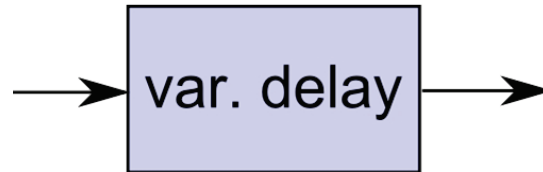
Echo

- Applications:
 - Repetition of sound event
 - Simulation of an early reflection added to direct sound
 - Broadening and softening a sound



Vibrato

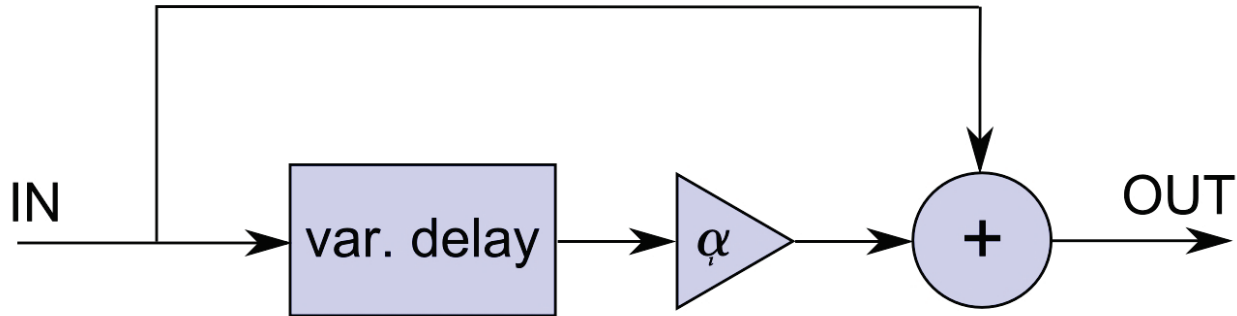
- Typical characteristic in singing voice, violin or guitar
- Time-variable pitch
 - Modulation frequency of a few Hz
 - Modulation depth: below semitone = below 6%
- Implementation by variable delay steered by oscillator, typically sinusoid
- Time-dependence → no LTI system



Flanger

Flanger: Principle

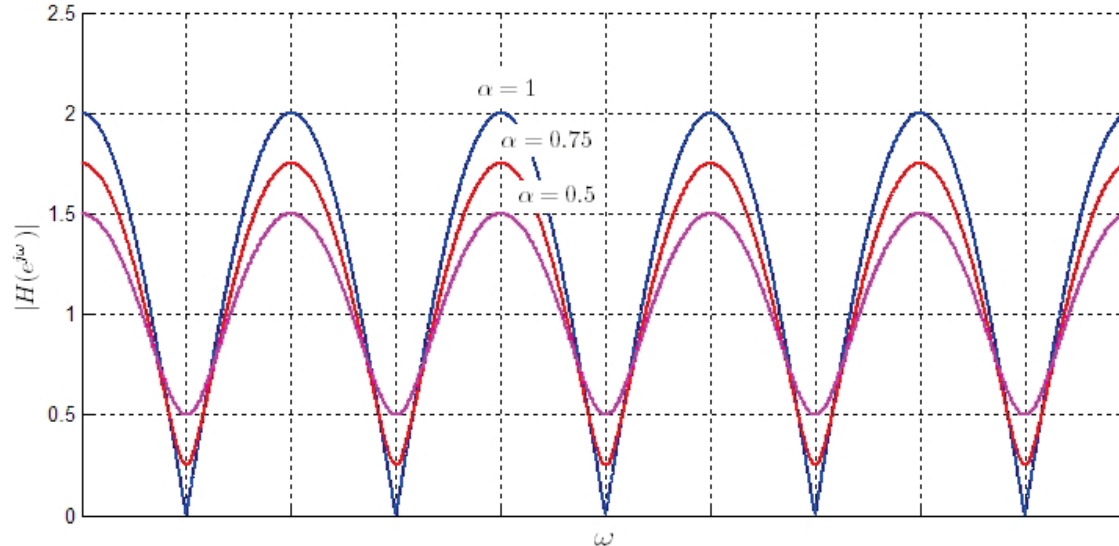
- summation of signal and a delayed copy
- slow variation (0.1 Hz) of the delay (couple of milliseconds)



Flanger: Comb filter

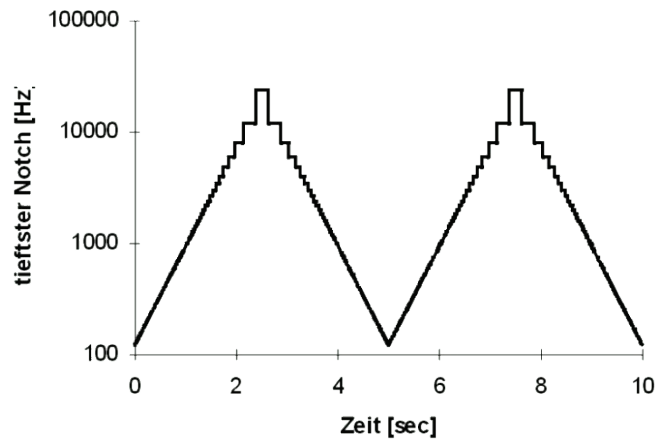
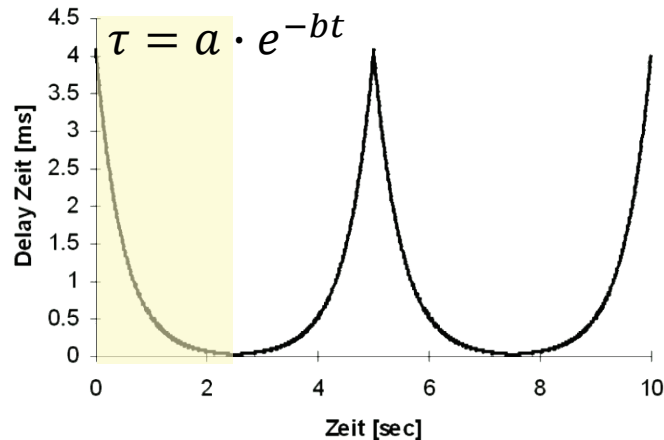
- Comb filter for delay τ :

- Maxima: $f = 0, \frac{1}{\tau}, \frac{2}{\tau}, \frac{3}{\tau}$ Minima: $f = \frac{0.5}{\tau}, \frac{1.5}{\tau}, \frac{2.5}{\tau}$





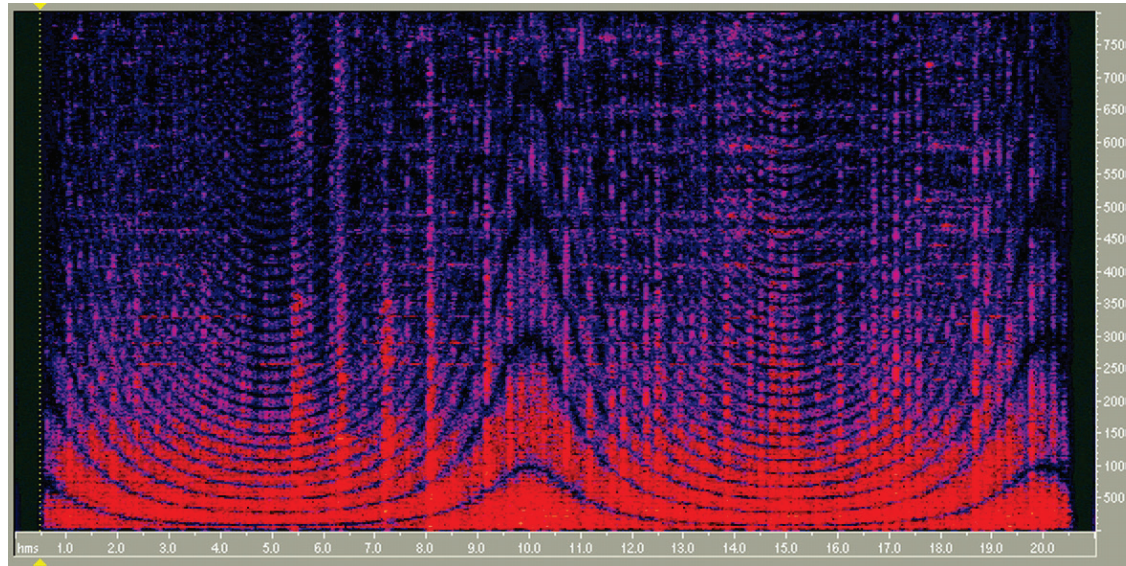
Flanger: Comb filter

- time dependency of the delay variation function
 - Different variants
 - e.g. sinusoidal: uneven perceived variation of the maxima/minima frequencies due to logarithmic frequency perception
 - in general → non-sinusoidal time function



Flanger: Example

- Original sample: 
- Flanger with delay: 0.5-5ms, variation: sinusoidal with 0.1Hz: 



Chorus

- Application: Virtual multiplication of an instrument or voice
- Parallel arrangement of independent flangers

■ Original: 

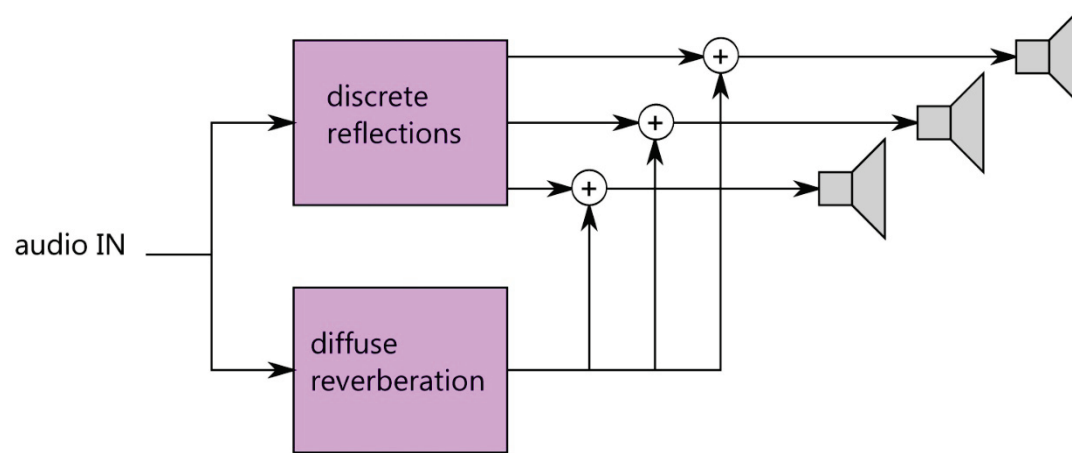
■ With chorus: 

Phaser

- Similar to flanger but summation of signal and *phase-shifted* copy
- Slow variation of the phase shift
- Result = Spectral - non-harmonic - amplifications and attenuations
- Realization by cascade of time-varying all-pass filters

Reverberation

Reverberation: Room simulation



Reverberation: Realization of diffuse reverberation

- Analog
 - Spring reverb: mechanical springs
 - Plate reverb: vibration of metal plate
 - Echo chambers: reverberant acoustical space → KKL
- Digital
 - Feed-back structures (IIR)
 - Convolution with impulse responses (FIR)



Artificial reverberation

- Many reverb algorithms
- Original work by M. Schroeder

[Schroeder, M.R. 1962. Natural sounding artificial reverberation. *Journal of the Audio Engineering Society*, 10(3).]

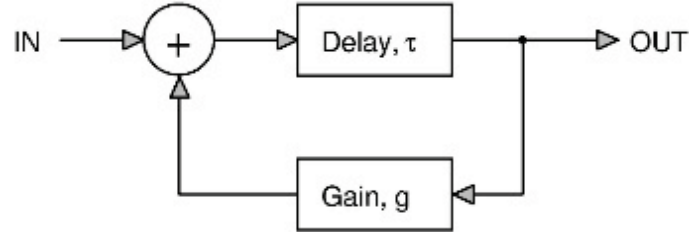
[Välimäki, Vesa et al. 2012. Fifty Year of Artificial Reverberation. *IEEE Transactions on Audio, Speech and Language Processing* 20(5).

Smith III, Julius O. 2010. *Physical Audio Signal Processing: for Virtual Musical Instruments and Digital Audio Effects*. W3K Publishing.]

<https://valhalladsp.com/2009/05/30/schroeder-reverbs-the-forgotten-algorithm/>
[https://ccrma.stanford.edu/~jos/pasp/Artificial Reverberation.html](https://ccrma.stanford.edu/~jos/pasp/Artificial_Reverberation.html)

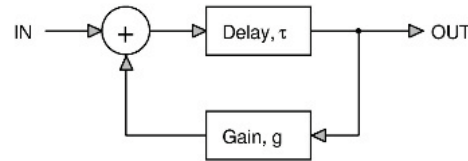
Artificial reverberator element

- Simplest feedback structure

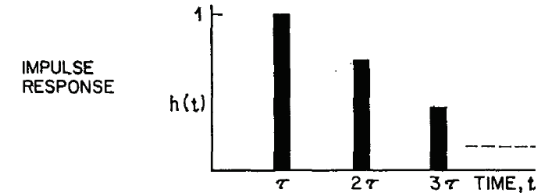


- Condition for stability: $g < 1$
- Physical concept: Plane wave travelling between two parallel walls with slight absorption \rightarrow flutter echo

Artificial reverberator element



- Impulse response:



$$h(t) = \delta(t - \tau) + g\delta(t - 2\tau) + g^2\delta(t - 3\tau) + \dots$$

- Transfer function:

$$H(\omega) = e^{-j\omega\tau} + ge^{-j\omega 2\tau} + g^2e^{-j\omega 3\tau} + \dots$$

Artificial reverberator element

$$H(\omega) = e^{-j\omega\tau} + ge^{-j\omega 2\tau} + g^2 e^{-j\omega 3\tau} + \dots$$

is a geometric series \rightarrow

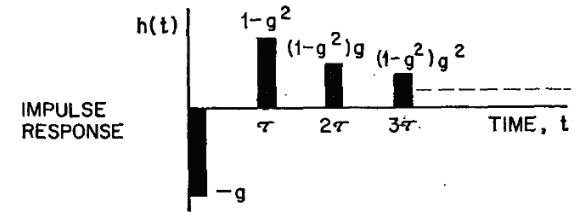
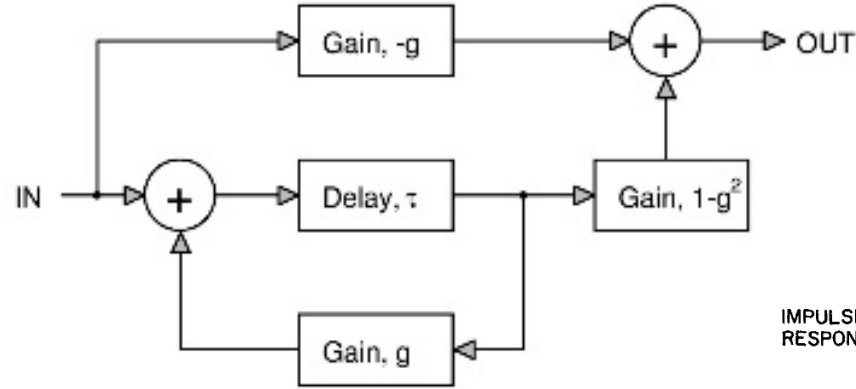
$$H(\omega) = \frac{e^{-j\omega\tau}}{1 - ge^{-j\omega\tau}}$$

with magnitude response:

$$|H(\omega)| = \frac{1}{\sqrt{1 + g^2 - 2g \cos \omega\tau}}$$

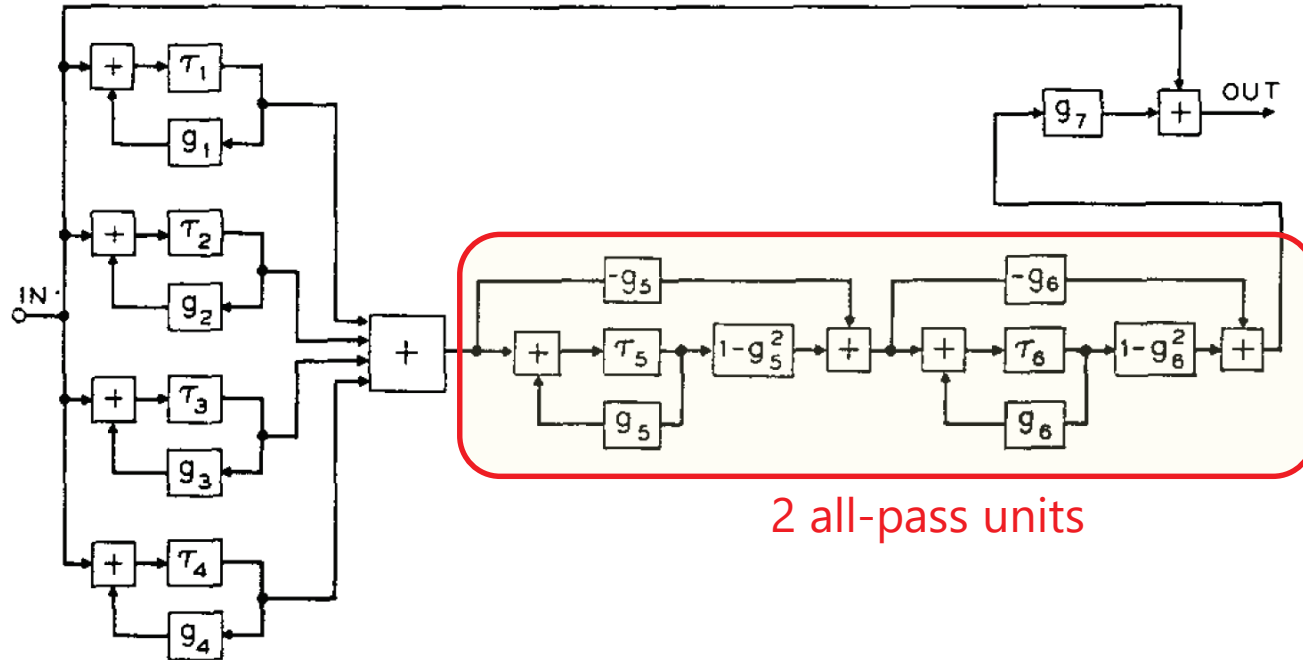
\rightarrow comb filter function! «feedback comb filter»

Artificial reverberator element: All-pass



- Feedback + feed-forward comb filter = All-pass
- Flat amplitude response
- Reverberation time: $RT = \frac{60\tau}{-20\log(g)}$
- Issue: For $g \rightarrow 1$ direct sound path dominates
- Use of multiple elements \rightarrow Schroeder

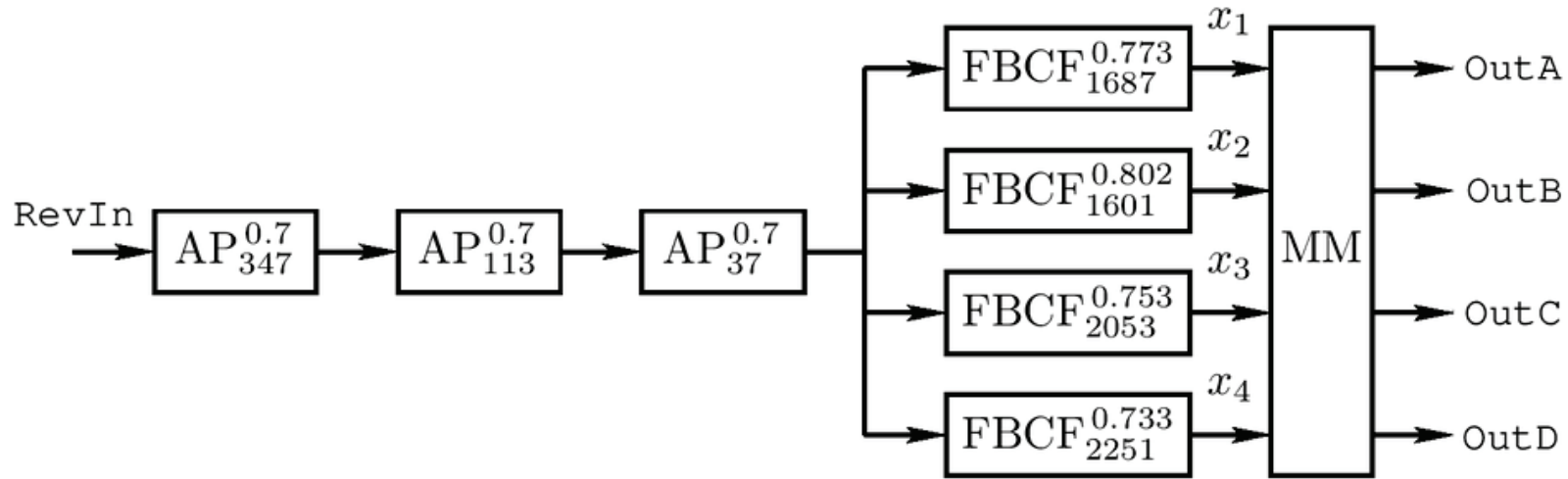
Schroeder reverberator 1962



2 all-pass units

4 feedback comb filter units

Schroeder type reverberators



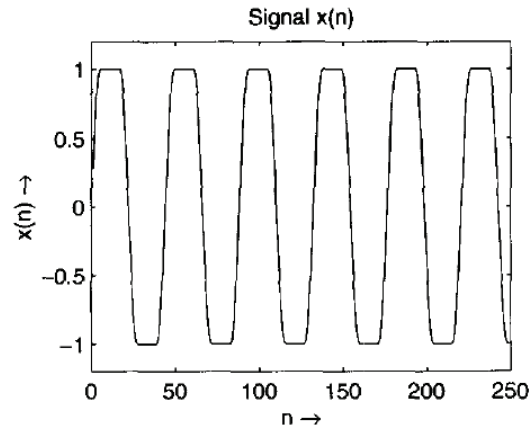
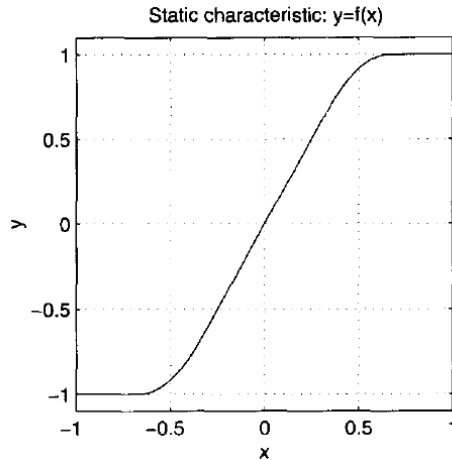
- FBCF: Feedback comb filter
- AP: All-pass = feedback + feedforward comb filter
- MM: Matrix mixing

Convolutional Reverb

- Reverberation is linear effect
- Simulation by linear convolution with room impulse response (RIR)
- RIR:
 - Truncation to get finite length
 - may be quite long: up to 15 seconds
 - much shorter than the input signal
- Digital implementation: FIR filter
- Efficient frequency domain algorithm using FFT/iFFT and overlap-add method

Other effects

Overdrive, distortion

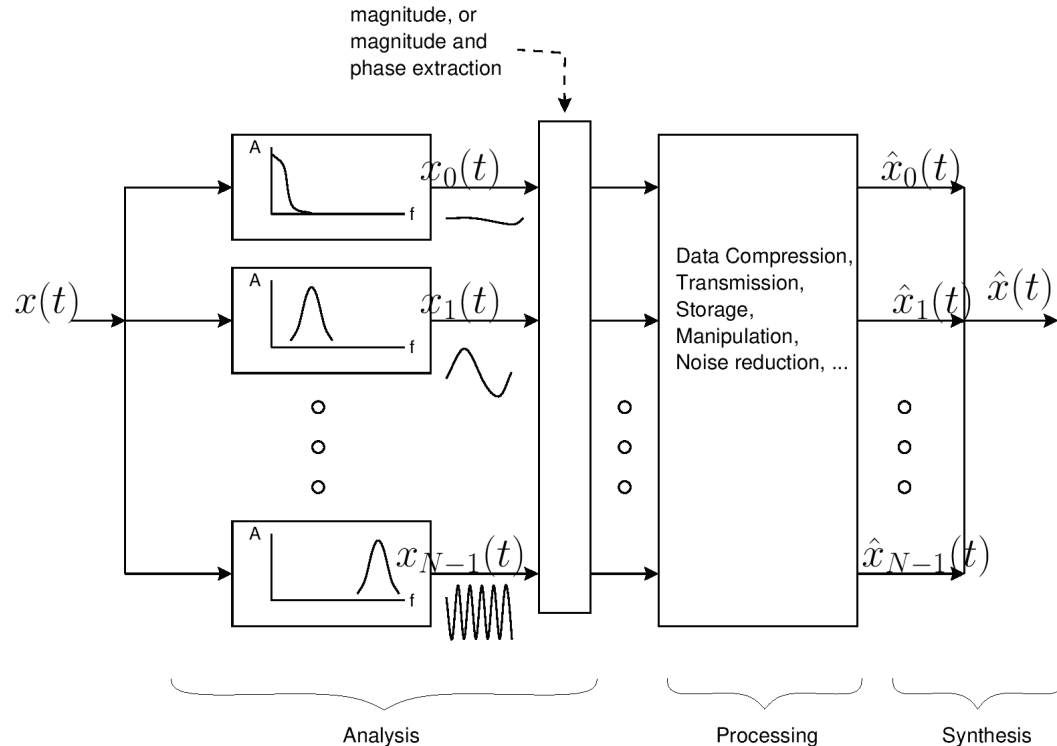


- Symmetrical clipping: creation of odd harmonics only
- Asymmetrical clipping: creation of even and odd harmonic

[Zölzer, Udo. et al. 2002. DAFX – Digital Audio Effects. John Wiley & Sons.]

Vocoder

Vocoder = Voice coder: Principle



Vocoder Example:
Daft Punk – Get Lucky







- Channel vocoder: Energy in bands
- Phase vocoder: Considers phase through STFT

Pitch-shift and time scaling

Pitch-shift and time scaling

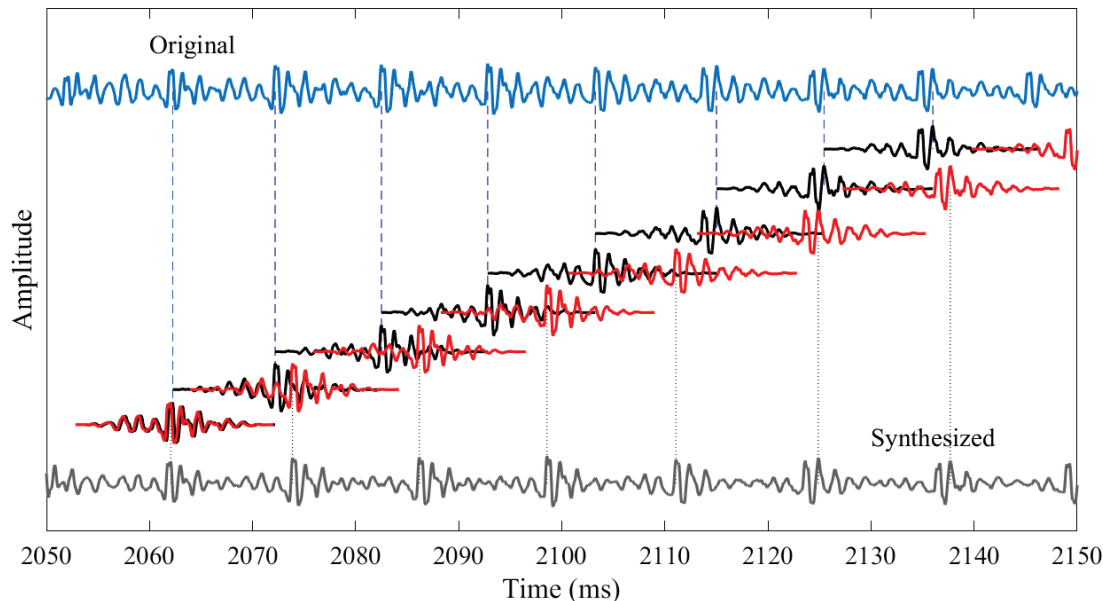
- Goal
 - Pitch variation without scaling time axis
 - Time axis scaling without pitch shift
- Examples:

	Original	Stretching factor 2
Violin		
Drums		

from Průša 2017, <https://lftfat.github.io/notes/050/>

Pitch-shift and time scaling : PSOLA

- Time-domain approach: Pitch-Synchronous OverLap and Add
- Processing: Pitch estimation, segmentation, windowing, assembling



[<https://wiki.aalto.fi/display/ITSP/Introduction+to+Speech+Processing>]

Pitch-shift and time scaling : Phase vocoder

- Frequency-domain method using the Short-Time Fourier Transform (STFT)
- Usual problem: Artefacts like phasiness or loss of presence
- Many algorithms for consideration of phase coherence across time and frequency → phase locking methods

[Laroche, J. & Dolson, M. 1999. Improved Phase Vocoder Time-Scale Modification of Audio. IEEE Transactions on Speech and Audio Processing, 7(3).]