### Control of Low-Inertia Power Systems: Naive & Foundational Approaches

(extended set of slides)

Florian Dörfler



### Acknowledgements













B.K. Poolla C. Arghir

C. Arghir T. Jouini

ouini P. Lütolf

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D. Groß

3/54



S. Bolognani



S. Curi



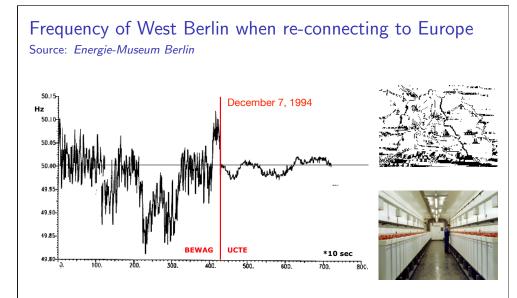


HORIZ N 2020

M. Colombino

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## What do we see here?



**before** re-connection: islanded operation based on batteries & single boiler **afterwards** connected to European grid based on synchronous generation

### Essentially, the pre/post West Berlin curves date back to...

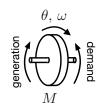


Fact: all of AC power systems built around synchronous machines!

At the heart of it is the generator swing equation:

$$M \frac{d}{dt} \omega(t) = P_{\text{generation}}(t) - P_{\text{demand}}(t)$$

change of kinetic energy = instantaneous power balance



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## Operation centered around bulk synchronous generation Frimary Control Primary Control Secondary Control Oscillation/Control 49.99 49.91 49.91 49.91 49.91 49.91 49.91 49.91 49.91 49.92 49.91 Frequency Mettlen, Switzerland

### Renewable/distributed/non-rotational generation on the rise

synchronous generator

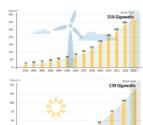


new workhorse



scaling

new primary sources



location & distributed implementation



focus today on **non-rotational** generation

### The foundation of today's power system







Source: W. Sattinger, Swissgrid

Synchronous machines with rotational inertia

$$M\frac{d}{dt}\omega \approx P_{\text{generation}} - P_{\text{demand}}$$

Today's grid operation heavily relies on

- 1 robust stabilization of frequency and voltage by generator controls
- 2 self-synchronization of machines through the grid
- **3** kinetic energy  $\frac{1}{2}M\omega^2$  as **safeguard** against disturbances

We are replacing this solid foundation with . . .

### Tomorrow's clean and sustainable power system







Non-synchronous generation connected via power electronics

As of today, power electronic converters

- lack robust control for voltage and frequency
- 4 do not inherently synchronize through the grid
- provide almost no energy storage

What could possibly go wrong?

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### Black System Event in South Australia (Sep 2016)

The Sydney Morning Merald

NATIONAL

State in the dark: South Australia's major power outage

south Australia
South Australia blackout: entire state left without power after storms

### Key events<sup>1</sup>

- 1 intermittent voltage disturbances due to line faults
- 2 loss of synchronism between SA and remainder of the grid
- 3 SA islanded: frequency collapse in a quarter of a second

"Nine of the 13 wind farms online did not ride through the six voltage disturbances experienced during the event."

<sup>1</sup>AEMO: Update Report - Black System Event in South Australia on 28 September 2016

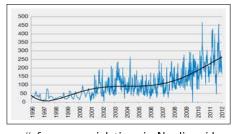
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### Low inertia issues have been broadly recognized

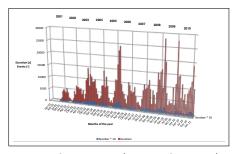
by TSOs, device manufacturers, academia, funding agencies, etc.



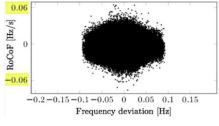
### Low-inertia issues close to home



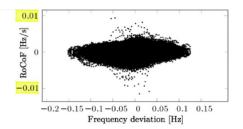
# frequency violations in Nordic grid (source: ENTSO-E)



same in Switzerland (source: Swissgrid)



a day in Ireland (source: F. Emiliano)



a year in France (source: RTE)

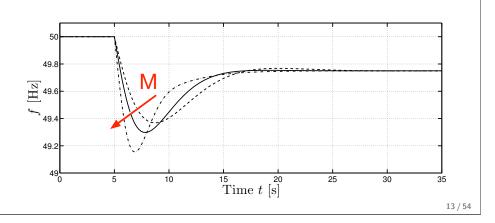
### Obvious insight: loss of inertia & frequency stability

We loose our giant electromechanical low-pass filter:

$$\mathbf{M} \frac{d}{dt} \omega(t) = P_{\text{generation}}(t) - P_{\text{demand}}(t)$$

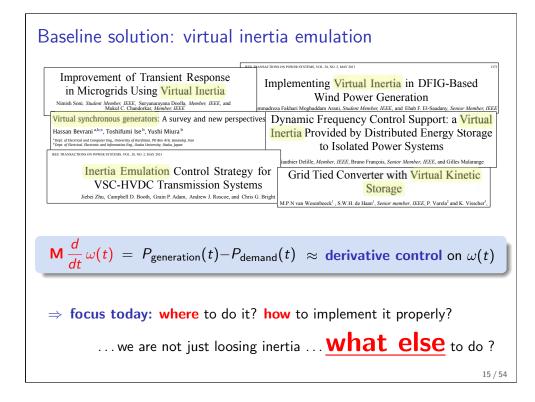
 $\theta, \omega$ defination

change of kinetic energy = instantaneous power balance



# Berlin curves before and after re-connecting to Europe Source: Energie-Museum Berlin 50,0 Hz 49,8 loss of 1200 MW Berlin re-connected to Europe 149,6 49,4 49,0 Islanded Berlin grid loss of 146 MW

## obvious insights lead to obvious (naive) answers



### Outline

Introduction

System Level: Optimal Placement of Virtual Inertia network, disturbances, & performance metrics matter

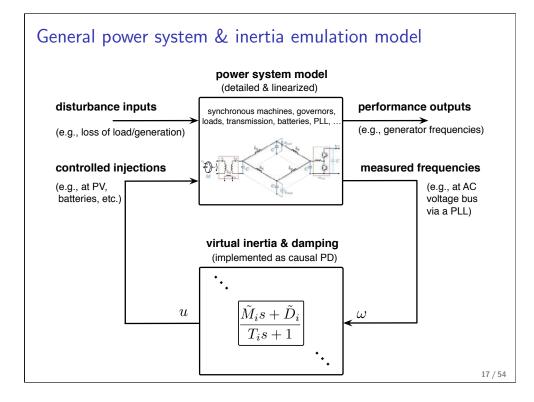
Device Level: Proper Virtual Inertia Emulation Strategy maybe we should not think about frequency and inertia

A Foundational Control Approach restart from scratch for low-inertia systems

**Conclusions** 

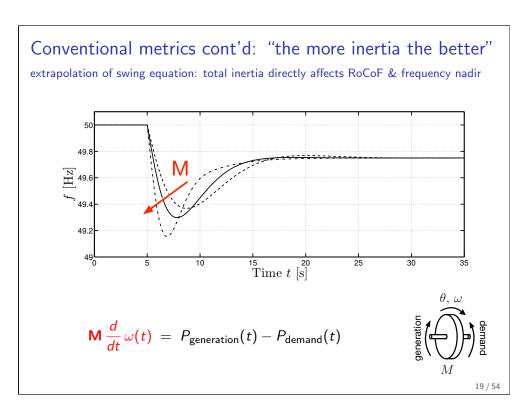
optimal placement of virtual inertia

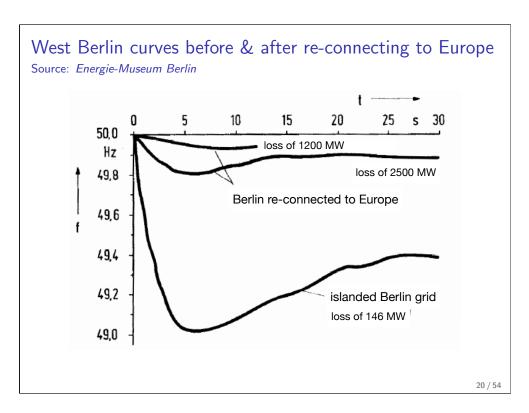




### which metric(s) should our controller optimize?

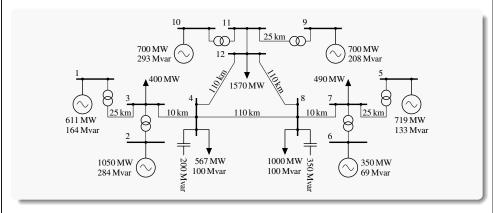
### Conventional metrics disturbance inputs: performance outputs: • step (loss of load/generation) overshoot (peak signals after fault) • RoCoF (rate of change of frequency) • impulse (line open-/closing) • noise (renewables & loads) • spectrum (damping ratio cones) re-evaluate scenario? hardly tractable for optimization & control design ROCOF (max rate of change of frequency) metrics & faults justified only in a post-fault response in a low-inertia system? system dominated by machines metrics any useful?



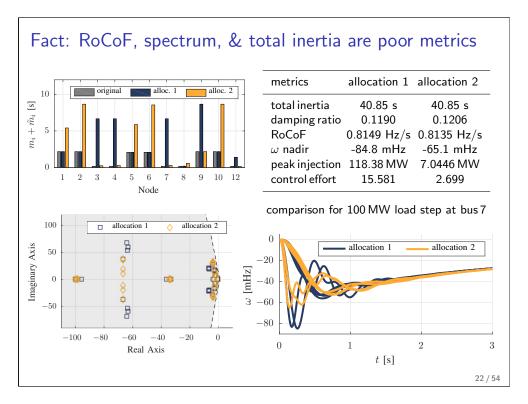


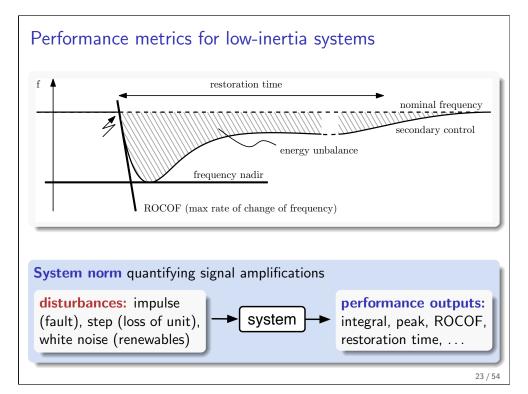
## are these suitable metrics? let's look at some simulations

### Running example: modified Kundur three-area case study



- added third area to standard case
- PLLs at all buses for inertia emulation (overall device response time ∼100ms)
- transformer reactance 0.15 p.u, line impedance (0.0001+0.001i) p.u./km
- original inertia 40s: removed of rotational 28s which can be re-allocated as virtual inertia
- added governors & droop control at all generators





### Integral-quadratic coherency performance metric

$$\int_0^\infty x(t)^T Q x(t) dt$$



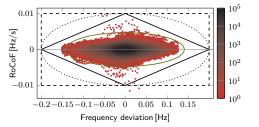
 $\mathcal{H}_2$  system norm interpretation:  $\eta \longrightarrow system \longrightarrow \mathcal{I}$ 

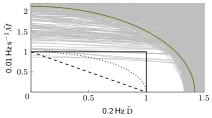
- **1** performance output:  $y = Q^{1/2}x$
- 2 impulsive  $\eta$  (faults)  $\longrightarrow$  output energy  $\int_0^\infty \mathbf{y}(t)^\mathsf{T} \mathbf{y}(t) dt$
- **3** white noise  $\eta$  (renewables)  $\longrightarrow$  output variance  $\lim_{t \to \infty} \mathbb{E}\left(\mathbf{y}(t)^\mathsf{T} \mathbf{y}(t)\right)$

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### Constraints on control inputs

- **1** energy constraint:  $\int_0^\infty u^T R u \, dt$  directly captured in  $\mathcal{H}_2$  framework
- **2** power constraint:  $u_i = \tilde{M}_i \dot{\omega}_i + \tilde{D}_i \omega_i$  must satisfy  $||u_i(t)||_{\ell_{\infty}} \leq \overline{u_i}$





European frequency data (source: RTE)

corresponding bounds on gains

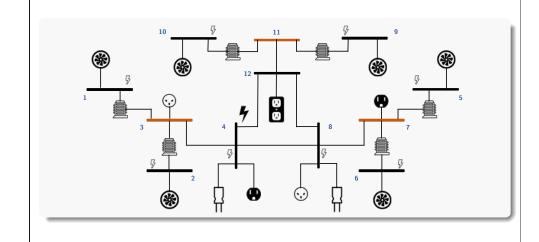
- $\Rightarrow \|(\omega_i(t),\dot{\omega}_i(t))\|_p$ ,  $\|( ilde{D}_i, ilde{M}_i)\|_q$  bounded  $(rac{1}{p}+rac{1}{q}=1) \Rightarrow \|u_i(t)\|_{\ell_\infty}$  bounded
- **3 budget constraint** for finitely many devices:  $\sum_{i} \overline{u_i} = const.$

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### (sub)optimize performance and see what we learn

### Modified Kundur case study: 3 areas & 12 buses

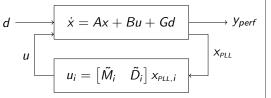
added governors (droop) at generators & PLLs to obtain frequency for inertia emulation



### Test case

• inertia emulation control via PLL & batteries:

$$u_i = \begin{bmatrix} \tilde{M}_i & \tilde{D}_i \end{bmatrix} x_{\scriptscriptstyle PLL,i}$$



• dynamics: swing equation, droop via governor & turbine, and PLL

$$\begin{bmatrix} \dot{\delta} \\ \dot{\omega} \\ \dot{x}_{gov} \\ \dot{x}_{PLL} \end{bmatrix} = \underbrace{\begin{bmatrix} A_{sw} & B_{sw} K_{gov} & \mathbb{O} \\ B_{gov} & A_{gov} & \mathbb{O} \\ B_{PLL} & \mathbb{O} & A_{PLL} \end{bmatrix}}_{=A} x + \underbrace{\begin{bmatrix} B_{sw} \\ \mathbb{O} \\ \mathbb{O} \end{bmatrix}}_{=B} u + \underbrace{\begin{bmatrix} B_{sw} \\ \mathbb{O} \\ \mathbb{O} \end{bmatrix}}_{=G} d$$

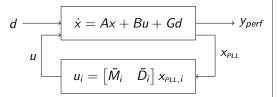
 cost penalizes frequencies, droop control, & inertia emulation effort:

$$\underbrace{\begin{bmatrix} \omega \\ u_{gov} \\ u \end{bmatrix}}_{y_{perf}} = \underbrace{\begin{bmatrix} 0 & I & 0 & 0 \\ 0 & 0 & K_{gov} & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}}_{=Q^{1/2}} x + \underbrace{\begin{bmatrix} 0 \\ 0 \\ I \end{bmatrix}}_{=R^{1/2}} u$$

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### Algorithmic approach to desperate & non-convex problem

 structured state-feedback with constraints on gains



• **computation**  $\mathcal{H}_2$  norm, gradient, & projections:

• observability and controllability Gramians via Lyapunov equations

$$(A - BK)^{\top}P + P(A - BK) + Q + K^{\top}RK = 0$$
$$(A - BK)L + L(A - BK)^{\top} + GG^{\top} = 0$$

- **2**  $\mathcal{H}_2$  norm  $J = \text{Trace}(G^{\top}PG)$  and gradient  $\nabla_K J = 2(RK B^{\top}P)L$
- **3** projection on structural &  $\infty$ -norm constraint:  $\Pi_{\tilde{M},\tilde{D}}[\nabla_K J]$
- $\Rightarrow \tilde{M}$  and  $\tilde{D}$  can be optimized by first-order methods, IPM, SQP, etc.

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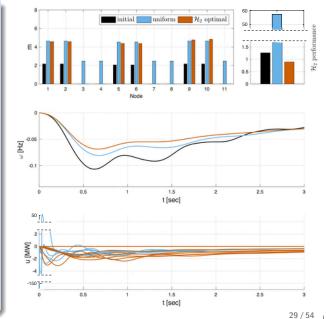
### Results & insights for the three-area case study

### **Optimal allocation:**

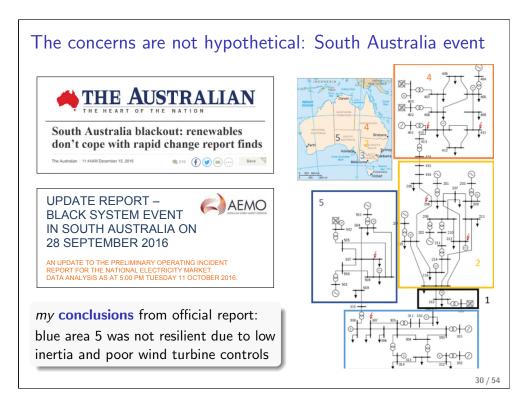
- location of inertia & damping matters
- outperforms heuristic uniform allocation
- need penalty on droop control effort
- power constraint results in  $\tilde{D} \approx 2\tilde{M}$

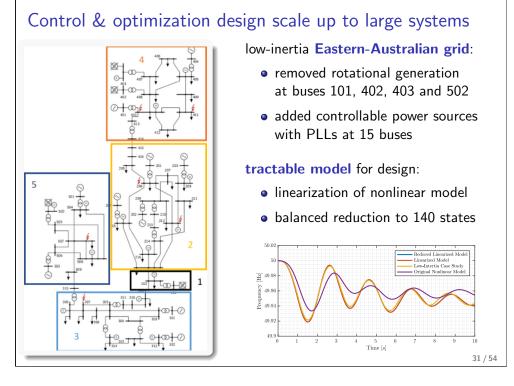
### Fault at bus #4:

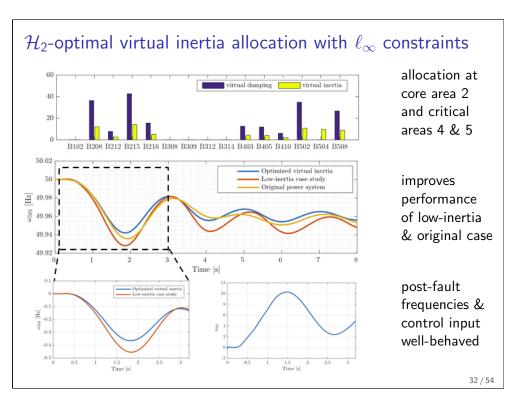
- strong reduction of frequency deviation
- much less control effort than heuristic



## can we make this control design strategy useful?







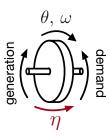
placement & metrics matter! can we get analytic insights?

### Inertia placement in swing equations

• simplified network swing equation model:

$$m_i \ddot{\theta}_i + d_i \dot{\theta}_i = p_{gen,i} - p_{dem,i}$$
  
generator swing equations

$$p_{dem,i} \approx \sum_{j} b_{ij} (\theta_i - \theta_j)$$
  
linearized DC power flow



- likelihood of disturbance at #i:  $\eta_i \ge 0$  (available from TSO data)
- $\mathcal{H}_2$  performance **metric**:  $\int_0^\infty \sum_{i,j} a_{ij} (\theta_i \theta_j)^2 + \sum_i s_i \dot{\theta}_i^2 dt$
- decision variable is inertia:  $m_i \in [\underline{m_i}, \overline{m_i}]$  (additional nonlinearity: enters as  $m_i^{-1}$  in constraints & objective)  $_{33/54}$

### Closed-form results for cost of primary control

recall: primary control  $d_i \dot{\theta}_i$  effort was crucial

$$\int_0^\infty \dot{\theta}(t)^\mathsf{T} D \,\dot{\theta}(t) \,dt$$

(computations show that insights roughly generalize to other costs)

**allocation:** the primary control effort  $\mathcal{H}_2$  optimization reads equivalently as

minimize 
$$\sum_{i} \frac{\eta_i}{m_i}$$

subject to 
$$\sum_i m_i \leq m_{\text{bdg}}$$

$$m_i \leq m_i \leq \overline{m_i}$$

key take-away is disturbance matching:

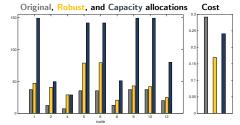
- optimal allocation  $m_i^{\star} \propto \sqrt{\eta_i}$  or  $m_i^{\star} = \min\{m_{\text{bdg}}, \overline{m_i}\}$
- ⇒ disturbance profile known from historic data, but rare events are crucial
- ightharpoonup suggests robust min<sub>m</sub> max<sub>n</sub> allocation to prepare for worst case
- $\Rightarrow$  valley-filling solution:  $\eta_i^{\star}/m_i^{\star} = const.$  (up to constraints)

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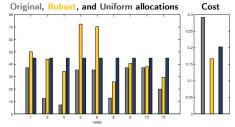
### Robust min-max allocation for three-area case study

**Scenario:** fault (impulse) can occur at any single node

- disturbance set  $\eta \in \{e_1 \cup \cdots \cup e_{12}\}$
- ⇒ min/max over convex hull
- ► inertia capacity constraints
- robust inertia allocation outperforms heuristic max-capacity allocation
- results become intuitive: valley-filling property
- ▶ same for uniform allocation



allocation subject to capacity constraints



allocation subject to the budget constraint

Outline

Introduction

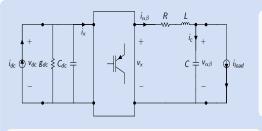
System Level: Optimal Placement of Virtual Inertia network, disturbances, & performance metrics matter

Device Level: Proper Virtual Inertia Emulation Strategy maybe we should not think about frequency and inertia

A Foundational Control Approach restart from scratch for low-inertia systems

Conclusions

### Averaged power converter model



### DC cap & AC filter equations:

$$C_{dc}\dot{v}_{dc} = -G_{dc}v_{dc} + i_{dc} - \frac{1}{2}m^{\top}i_{\alpha\beta}$$

$$Li_{\alpha\beta} = -Ri_{\alpha\beta} + \frac{1}{2}mv_{dc} - v_{\alpha\beta}$$

$$C\dot{v}_{\alpha\beta} = -i_{load} + i_{\alpha\beta}$$

modulation: 
$$v_{x} = \frac{1}{2} m v_{dc}$$
,  $i_{x} = \frac{1}{2} m^{\top} i_{\alpha\beta}$ 

control/dist. inputs:  $(i_{dc}, i_{load})$ 

### synchronous generator: mechanical + stator flux

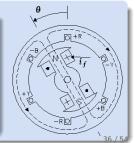
+ AC cap

$$\dot{\theta} = \omega$$

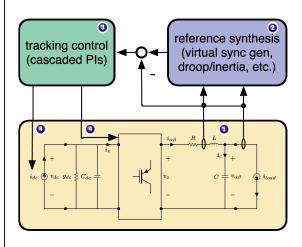
$$M\dot{\omega} = -D\omega + \tau_m + i_{\alpha\beta}^{\mathsf{T}} L_m i_f \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$

$$L_s i_{\alpha\beta} = -R i_{\alpha\beta} - v_{\alpha\beta} - \omega L_m i_f \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$

$$C\dot{v}_{\alpha\beta} = -i_{load} + i_{\alpha\beta}$$



### Standard power electronics control would continue by



- acquiring & processing of AC measurements
- 2 synthesis of references (voltage/current/power)
- **1** track error signals at converter terminals
- actuation via modulation (inner loop) and/or via DC source (outer loop)

I guess you can see the problems building up ...

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### Challenges in power converter implementations



Real Time Simulation of a Power System with VSG Hardware in the Loop

- **1** delays in measurement acquisition, signal processing, & actuation
- 2 accuracy in AC measurements (averaging over multiple cycles)
- constraints on currents. voltages, power, etc.
- certificates on stability, robustness, & performance

### entso **Frequency Stability Evaluation** Criteria for the Synchronous Zone

of Continental Europe

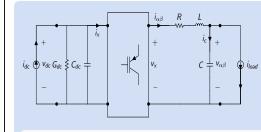
- Requirements and impacting factors -

RG-CE System Protection & Dynamics Sub Group

added to the inverter to provide "synthetic inertia". This can also be seen as a short term frequency support. On the other hand, these sources might be quite restricted with respect to the available capacity and possible activation time. The inverters have a very low

let's do something smarter . . .

### See the similarities & the differences?



DC cap & AC filter equations:

$$C_{dc}\dot{v}_{dc} = -G_{dc}v_{dc} + i_{dc} - \frac{1}{2}m^{\top}i_{\alpha\beta}$$
 $C_{dc}\dot{v}_{dc} = -Ri_{\alpha\beta} + \frac{1}{2}mv_{dc} - v_{\alpha\beta}$ 
 $C\dot{v}_{\alpha\beta} = -i_{load} + i_{\alpha\beta}$ 

modulation: 
$$v_x = \frac{1}{2} m v_{dc}$$
,  $i_x = \frac{1}{2} m^{\top} i_{\alpha\beta}$ 

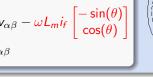
passive:  $(i_{dc}, i_{load}) \rightarrow (v_{dc}, v_{\alpha\beta})$ 

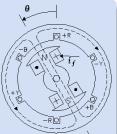
synchronous generator: mechanical + stator flux + AC cap

$$\theta = \omega$$

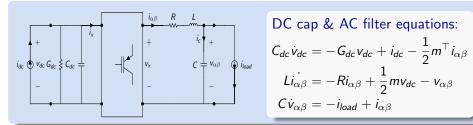
$$M\dot{\omega} = -D\omega + \tau_m + i_{\alpha\beta}^{\top} L_m i_f \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$

$$L_{s}i_{\alpha\beta}^{\cdot} = -Ri_{\alpha\beta} - v_{\alpha\beta} - \omega L_{m}i_{f}\begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$
  
 $C\dot{v}_{\alpha\beta} = -i_{load} + i_{\alpha\beta}$ 





### Model matching ( $\neq$ emulation) as inner control loop



### DC cap & AC filter equations:

$$C_{dc}\dot{v}_{dc} = -G_{dc}v_{dc} + i_{dc} - \frac{1}{2}m^{\top}i_{\alpha\beta}$$

$$Li_{\alpha\beta}^{\cdot} = -Ri_{\alpha\beta} + \frac{1}{2}mv_{dc} - v_{\alpha\beta}$$

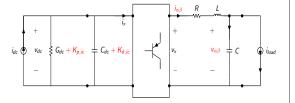
$$C\dot{v}_{\alpha\beta} = -i_{load} + i_{\alpha\beta}$$

**matching control**: 
$$\dot{\theta} = K_m \cdot v_{dc}$$
,  $m = \hat{m} \cdot \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$  with  $K_m$ ,  $\hat{m} > 0$ 

- $\Rightarrow$  equivalent inertia  $M=\frac{C_{dc}}{K_m^2}$ , droop/dissipation  $D=\frac{G_{dc}}{K_m^2}$ , torque  $au_m = rac{i_{dc}}{K_m}$ , field current  $i_f = rac{\hat{m}}{K_m L_m}$ , & imbalance signal  $\omega = K_m \cdot v_{dc}$
- ⇒ pros: uses physical storage, uses DC measurements, & remains passive

### Further properties of machine matching control

- base for outer loops
- $\Rightarrow i_{dc} = PD(v_{dc})$  gives virtual inertia & damping

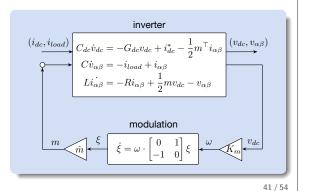


- droop slopes & nose curves. & further outer loops  $\hat{m}(\|v_{\alpha\beta}\|)$
- reformulation of

$$m = \hat{m} \cdot \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$

as adaptive **oscillator**:

$$\dot{m} = K_m \, v_{dc} \cdot \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix} m$$



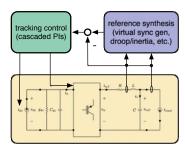
### Summary: bottlenecks to inertia emulation

power system model on grid level:

$$M\frac{d}{dt}\omega = P_{\text{generation}} - P_{\text{demand}}$$



inertia emulation on device level:



- I/O mismatch: none of the converter inputs or outputs are present in the swing-equation, e.g., frequency is not a state in the converter
- inertia emulation à la PD problematic both in theory & practice
- $\Rightarrow$  maybe matching control  $\dot{m} = K_m v_{dc} \cdot \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix} m$  was quite clever?

### Outline

Introduction

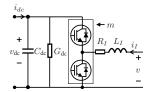
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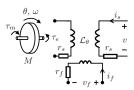
**Device Level: Proper Virtual Inertia Emulation Strategy** maybe we should not think about frequency and inertia

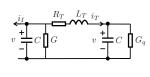
A Foundational Control Approach restart from scratch for low-inertia systems

**Conclusions** 

### Low-inertia power system model from first principles







- balanced three-phase system
  - $(\alpha, \beta)$  coordinates
- synchronous machines
  - first principle, 5th order
- ► DC/AC inverters
  - averaged-switched
- ▶ nonlinear loads G(||v||)

- voltage bus charge dynamics
- dynamic transmission lines: Π-model

Port-Hamiltonian model

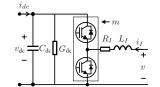
$$\dot{x} = \left(J(x, u) - R(x)\right) \nabla H(x) + g(x)u$$

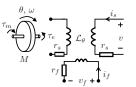
nonlinear & large, but insightful

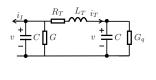
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### Desired steady-state locus & control specifications

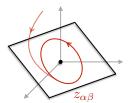






steady-state specifications for nonlinear system:

- synchronous frequency
- constant amplitude
- three-phase balanced



**AC** quantities  $v, i_s, i_l, i_T$ :

$$\dot{z}_{\alpha\beta} = \omega_0 \cdot \begin{bmatrix} 0 & -1 \\ 1 & 0 \end{bmatrix} z_{\alpha\beta}$$

rotor angles:  $\dot{\theta} = \omega_0$ 

**DC** quantities  $v_{dc}, v_f, \omega$ :  $\dot{z} = 0$ 

desired dynamics:  $\dot{x} = f_{des}(x, \omega_0)$ 

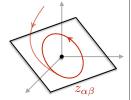
**controls**  $i_{dc}, m, \tau_m, i_f$  to be found

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### Proving the obvious (?)

 steady-state locus: physics & desired closed-loop vector field coincide (point-wise in time) on set

$$\mathcal{S} := \{(x, u, \omega_0) : f_{\mathsf{phys}}(x, u) = f_{\mathsf{des}}(x, \omega_0)\}$$



- control-invariance: steady-state operation  $(x, u, \omega_0) \in \mathcal{S}$  for all time if and only if
  - **1 synchronous frequency**  $\omega_0$  is constant
  - 2 **network** satisfies power flow equations with impedances  $R + \omega_0 JL$
  - **3** at each **generator**: constant torque  $\tau_m$  & excitation  $i_f$
  - **1** at each **inverter**: constant DC current  $i_{dc}$  & inverter duty cycle with constant amplitude & synchronous frequency:  $\dot{m} = \omega_0 \cdot \begin{bmatrix} 0 & -1 \\ 1 & 0 \end{bmatrix} m$
  - $\Rightarrow$  internal models & feedforward input-to-steady-state map

### Reduction to a tractable model for synthesis

• internal oscillator model for inverter duty cycle with inputs  $\omega_m$ ,  $\hat{m}$ 

$$\dot{\theta}_I = \omega_m, \quad m = \hat{m} \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$

- model reduction steps
  - **1 rotating coordinate frame** with synchronous frequency  $\omega_0$
  - $\Rightarrow$  time scales of AC quantities scaled by  $1/\omega_0$
  - **2** DC/AC time-scale separation via singular perturbation  $(\epsilon \to 0)$

slow DC variables:  $x_r = (\theta, \omega, i_f, \theta_I, v_{dc}),$   $\dot{x}_r = f_z(x_r, z_{\alpha,\beta}, u)$ 

fast AC variables:  $z_{\alpha,\beta} = (i_s, i_l, v, i_T),$   $\epsilon \dot{z}_{\alpha,\beta} = f_{\alpha,\beta}(x_r, z_{\alpha,\beta}, u)$ 

 $\odot$  reformulation via **relative angles**  $\delta$  with respect to synchronous motion

### Insights from reduced model: $v_{dc} \propto$ power imbalance

• nonlinear reduced order model in rotating frame:

$$\begin{split} \dot{\theta} &= \omega \\ M\dot{\omega} &= -D\omega + \tau_m - \tau_e(x_r, u) \\ L_f \dot{i}_f &= -R_f i_f + v_f - v_{EMF}(x_r, u) \\ \dot{\theta}_I &= \omega_m \\ C_{dc} \dot{v}_{dc} &= -G_{dc} v_{dc} + i_{dc} - i_{sw}(x_r, u) \end{split}$$

- interconnection via  $\tau_e$ ,  $i_{sw}$ ,  $v_{EMF}$
- analogies: suggest matching control:  $\omega_m \sim v_{dc}$

generator	inverter	interpretation
$rac{1}{2}M\omega^2$	$\frac{1}{2}C_{dc}v_{dc}^2$	energy stored in device
$ au_{m m}$	i <sub>dc</sub>	energy supply
$ au_{m{e}}$	i <sub>sw</sub>	energy flow to grid
$\omega$	V <sub>dc</sub>	power imbalance

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### Completing the control design

Thus far:

- desired steady-state locus requires internal oscillator model
- 2 converter/generator analogies suggest model matching control

### Remaining steps:

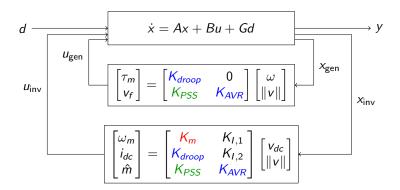
o robustness & stability under interconnection requires local feedback passification with respect to an incremental energy function

$$H_{\text{des}}(x) = \frac{1}{2}\omega^{T}M\omega + \frac{1}{2}(i_{f} - i_{f}^{*})^{T}L_{f}(i_{f} - i_{f}^{*}) + \frac{1}{2}(v_{dc} - v_{dc}^{*})^{T}C_{dc}(v_{dc} - v_{dc}^{*})^{T} + \dots$$

- $\Rightarrow$  associated passifying control is a scaled AC droop & DC droop
- performance requires design of structured & optimal MIMO control

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### Decentralized MIMO control architecture



- states  $x = (\delta, \omega, i_f, v_{dc}, ||v||)$  & output  $y = (\omega, v_{dc}, ||v||)$
- ullet included measurement devices for AC voltage magnitude  $\|v\|$
- H2-optimal tuning of decentralized MIMO converter controller

### Illustrative conceptual example

### test case:

- generator & inverter
- impedance load
- 10% load increase at t=0

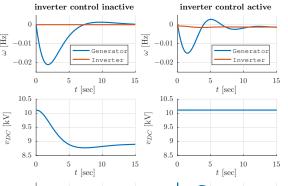
### no inverter control:

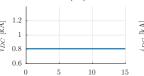
- $\omega_m$  and  $i_{dc}$  constant
- power imbalance:  $\omega_{\it G}$ ,  $v_{\it dc}$
- governor stabilizes  $\omega_G$

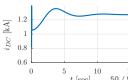
### controlled inverter:

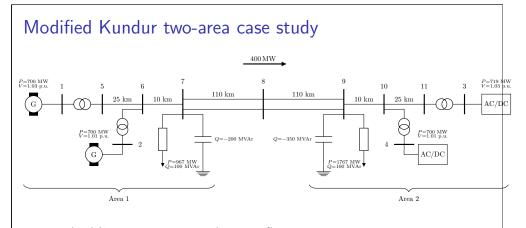
- reduced peak in  $\omega_G$
- $v_{dc}$  stabilized via  $i_{dc}$
- $\omega_m$  and  $\omega_G$  synchronize



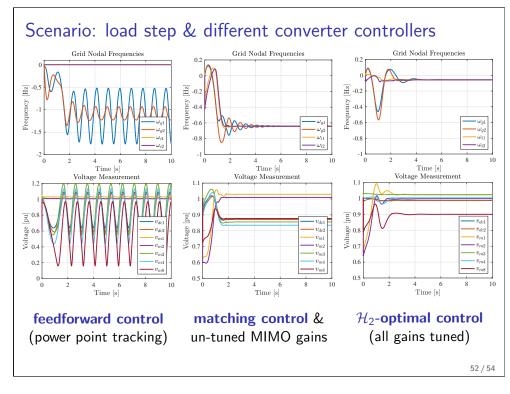


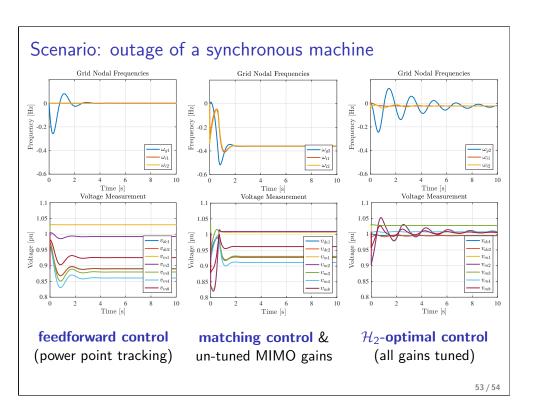






- standard line parameters and power flows
- synchronous machines with droop control and voltage regulator
- two synchronous machines replaced by DC/AC inverters
- all dirt effects modeled: saturation, nonlinearities, etc.
- simulation scenarios: load step  $(\times 2)$  & outage of synchronous machine







### Conclusions on virtual inertia emulation

### Where to do it?

- $\bullet$   $\mathcal{H}_2$ -optimal (non-convex) allocation
- 2 numerical approach via gradient computation
- 3 closed-form results for cost of primary control

### How to do it?

- down-sides of naive inertia emulation
- 2 machine matching reveals power imbalance in DC voltage

### What else to do?

- first-principle low-inertia system model
- nonlinear steady-state control specifications
- g reduction to tractable model for synthesis
- f 4 first promising controller synthesis: internal model + matching + passifying +  ${\cal H}_2$  performance loops

No power without control!

