Virtual Inertia Emulation and Placement in Power Grids

Séminaire d'Automatique du Plateau de Saclay Laboratoire de Signaux et Systèmes du Supelec

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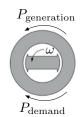
At the beginning of power systems was . . .



At the beginning was the synchronous machine:

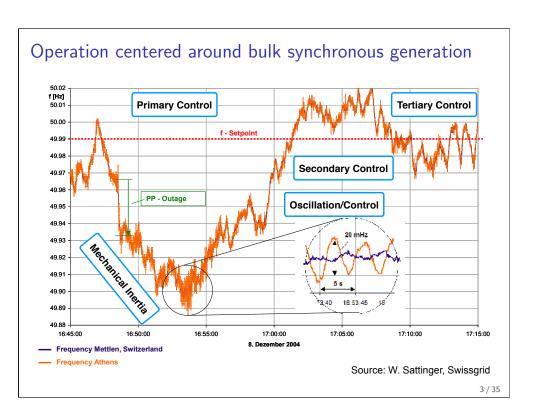
$$M \frac{d}{dt} \omega(t) = P_{\text{generation}}(t) - P_{\text{demand}}(t)$$

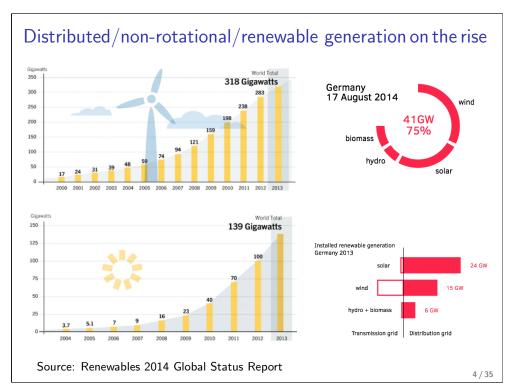
change of kinetic energy $\,=\,$ instantaneous power balance

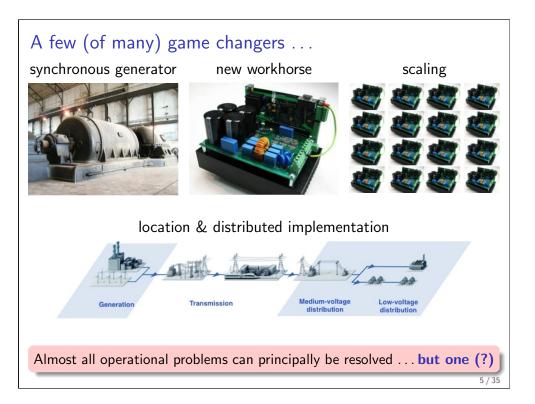


Fact: the AC grid & all of power system operation has been designed around synchronous machines.

2/35







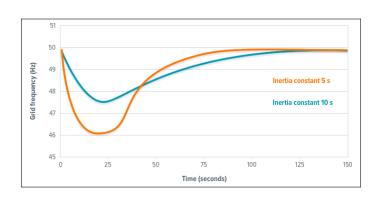
Fundamental challenge: operation of low-inertia systems

We slowly loose our giant electromechanical low-pass filter:

 $\mathbf{M} \frac{d}{dt} \omega(t) = P_{\text{generation}}(t) - P_{\text{demand}}(t)$

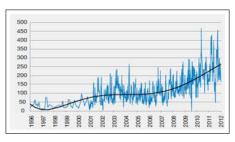
change of kinetic energy = instantaneous power balance



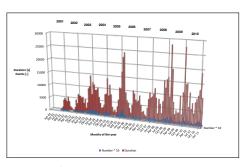


6 / 35

Low-inertia stability: $\#\ 1$ problem of distributed generation



frequency violations in Nordic grid (source: ENTSO-E)



same in Switzerland (source: Swissgrid)

inertia is shrinking, time-varying, & localized, ... & increasing disturbances

Solutions in sight: none really ... other than **emulating virtual inertia** through fly-wheels, batteries, super caps, HVDC, demand-response, ...

Virtual inertia emulation

devices commercially available, required by grid-codes or incentivized through markets

Improvement of Transient Response in Microgrids Using Virtual Inertia
Nimish Soni, Studem Member, IEEE, Suryanarayana Doolla, Member, IEEE, and Wind Power Generation
Nimish Soni, Studem Member, IEEE, Suryanarayana Doolla, Member, IEEE, and Wind Power Generation
Microgrids Using Virtual Inertia
Nimish Soni, Studem Member, IEEE, Suryanarayana Doolla, Member, IEEE, and Wind Power Generation
Microgrids Using Virtual Inertia
Wind Power Generation
Microgrids Wind

 $\mathbf{M} \frac{d}{dt} \omega(t) = P_{\text{generation}}(t) - P_{\text{demand}}(t)$... essentially **D-control**

- \Rightarrow plug-&-play (decentralized & passive), grid-friendly, user-friendly, \dots
- ⇒ today: where to do it? how to do it properly?

7 / 35

Outline

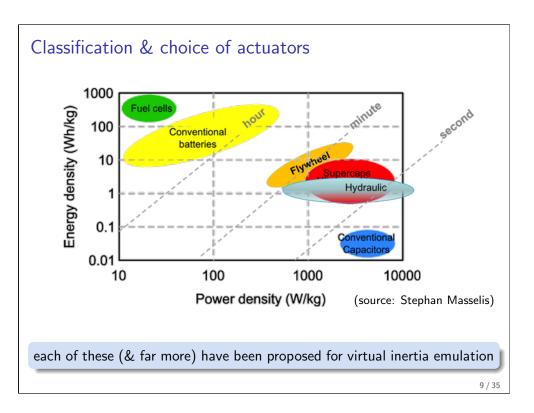
Introduction

Novel Virtual Inertia Emulation Strategy

Optimal Placement of Virtual Inertia

Conclusions

inertia emulation



Inertia emulation & virtual synchronous machines

a naive D-control on $\omega(t)$: $\mathbb{M} \frac{d}{dt} \omega(t) = P_{\text{generation}}(t) - P_{\text{demand}}(t)$ more sophisticated emulation of virtual synchronous machine

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Virtual synchronous generators: A survey and new perspectives
Hassan Bevrani $a^{h,e}$, Toshifumi Ise h , Yushi Miura h h -ppr, of thereised and Computer Eng. Under Other 415. Generally, Irea

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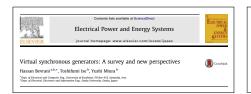
2- everything in between ... and much more ...

3- by measuring AC current/voltage/power/frequency

3- software model of virtual machine provides converter setpoints

⇒ actuation via modulation (switching) or DC injection (batteries etc.)

Challenges in real-world converter implementations

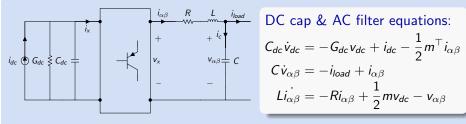


Real Time Simulation of a Power System with VSG Hardware in the Loop

- **1** delays in measurement acquisition, signal processing, & actuation
- 2 accuracy in AC measurements (averaged over ≈ 5 cycles)
- 3 constraints on currents, voltages, power, etc.
- guarantees on stability and robustness

today: use DC measurement, exploit analog storage, & passive control

Averaged inverter model



DC cap & AC filter equations:

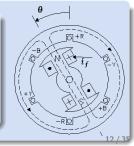
$$C_{dc}\dot{v}_{dc} = -G_{dc}v_{dc} + i_{dc} - \frac{1}{2}m^{\top}i_{\alpha\beta}$$
 $C\dot{v}_{\alpha\beta} = -i_{load} + i_{\alpha\beta}$
 $L\dot{i}_{\alpha\beta} = -R\dot{i}_{\alpha\beta} + \frac{1}{2}mv_{dc} - v_{\alpha\beta}$

modulation: $i_x = \frac{1}{2} m^{\top} i_{\alpha\beta}$, $v_x = \frac{1}{2} m v_{dc}$

passive: $(i_{dc}, i_{load}) \rightarrow (v_{dc}, v_{\alpha\beta})$

model of a synchronous generator

$$\begin{split} \dot{\theta} &= \omega \\ M\dot{\omega} &= -D\omega + \tau_m + i_{\alpha\beta}^{\top} L_m i_f \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix} \\ C\dot{v}_{\alpha\beta} &= -G_{load} v_{\alpha\beta} + i_{\alpha\beta} \\ L_s i_{\alpha\beta} &= -Ri_{\alpha\beta} - v_{\alpha\beta} - \omega L_m i_f \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix} \end{split}$$

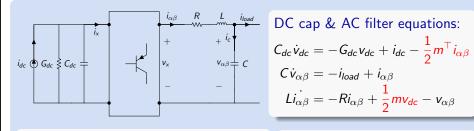


standard power electronics control would continue by

- constructing voltage/current/power references (e..g, droop, synchronous machine emulation, etc.)
- tracking these references at the converter terminals typically by means of cascaded PI controllers

let's do **something different** (smarter?) today ...

See the similarities & the differences?



$$C_{dc}\dot{v}_{dc} = -G_{dc}v_{dc} + i_{dc} - \frac{1}{2}m^{\top}i_{\alpha\beta}$$
 $C\dot{v}_{\alpha\beta} = -i_{load} + i_{\alpha\beta}$
 $Li_{\alpha\beta} = -Ri_{\alpha\beta} + \frac{1}{2}mv_{dc} - v_{\alpha\beta}$

modulation: $i_X = \frac{1}{2}m^{\top}i_{\alpha\beta}$, $v_X = \frac{1}{2}mv_{dc}$ passive: $(i_{dc}, i_{load}) \rightarrow (v_{dc}, v_{\alpha\beta})$

model of a synchronous generator

$$\theta = \omega$$

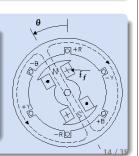
$$M\dot{\omega} = -D\omega + \tau_m + i_{\alpha\beta}^{\top} L_m i_f \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$

$$C\dot{v}_{\alpha\beta} = -G_{load} v_{\alpha\beta} + i_{\alpha\beta}$$

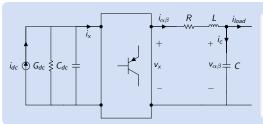
$$L_s i_{\alpha\beta} = -Ri_{\alpha\beta} - v_{\alpha\beta} - \omega L_m i_f \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$

$$C v_{\alpha\beta} = -G_{load} v_{\alpha\beta} + I_{\alpha\beta}$$

$$L_s \dot{i}_{\alpha\beta} = -R i_{\alpha\beta} - v_{\alpha\beta} - \omega L_m i_f \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$



Model matching (\neq emulation) as inner control loop



DC cap & AC filter equations:

$$C_{dc}\dot{v}_{dc} = -G_{dc}v_{dc} + i_{dc} - \frac{1}{2}m^{\top}i_{\alpha\beta}$$

$$C\dot{v}_{\alpha\beta} = -i_{load} + i_{\alpha\beta}$$

$$C\dot{v}_{\alpha\beta} = -Ri_{\alpha\beta} + \frac{1}{2}mv_{dc} - v_{\alpha\beta}$$

matching control:
$$\dot{\theta} = \eta \cdot \mathbf{v}_{dc}$$
, $m = \mu \cdot \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$ with $\eta, \mu > 0$

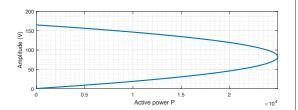
- \Rightarrow pros: is balanced, uses natural storage, & based on DC measurement
- \Rightarrow virtual machine with $M=rac{\mathcal{C}_{dc}}{\eta^2}$, $D=rac{\mathcal{C}_{dc}}{\eta^2}$, $au_m=rac{i_{dc}}{\eta}$, $i_f=rac{\mu}{\eta L_m}$
- \Rightarrow base for **outer controls** via i_{dc} & μ , e.g., virtual torque, PSS, & inertia

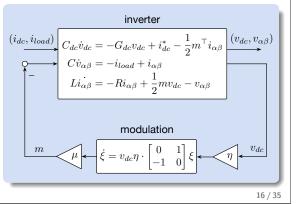
15 / 35

17/35

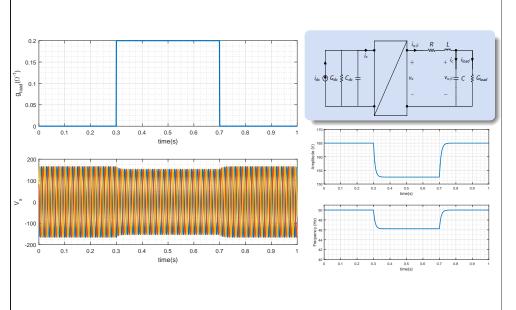
Some properties & different viewpoints

- quadratic curves for stationary P vs. $(|V|, \omega)$
- $\Rightarrow P \le P_{\text{max}} = i_{dc}^2 / 4G_{dc}$
- ⇒ reactive power not directly affected
- \Rightarrow (P, ω) -droop $\approx 1/\eta$
- \Rightarrow (*P*, |*V*|)-droop $\approx 1/\mu$
- reformulation as virtual & adaptive oscillator
- 3 remains passive: $(i_{dc}, i_{load}) \rightarrow (v_{dc}, v_{\alpha\beta})$





Eye candy: response to a load step



optimal placement of virtual inertia

Linearized & Kron-reduced swing equation model

$$m_i \ddot{\theta}_i + d_i \dot{\theta}_i = p_{in,i} - p_{e,i}$$

generator swing equations

$$p_{e,i} \approx \sum_{j \in \mathcal{N}} b_{ij} (\theta_i - \theta_j)$$
linearized power flows

 $P_{generation} + \eta$ P_{demand}

likelihood of **disturbance** at #i: $t_i \ge 0$

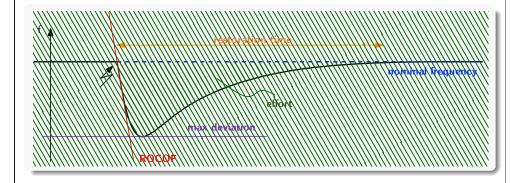
state space representation:

$$\begin{bmatrix} \dot{\theta} \\ \dot{\omega} \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & I \\ -M^{-1}L - M^{-1}D \end{bmatrix}}_{A} \begin{bmatrix} \theta \\ \omega \end{bmatrix} + \underbrace{\begin{bmatrix} 0 \\ M^{-1} \end{bmatrix}}_{B} T^{1/2} \eta$$

where $M = \operatorname{diag}(m_i)$, $D = \operatorname{diag}(d_i)$, $T = \operatorname{diag}(t_i)$, & $L = L^T$ (Laplacian)

18 / 35

Performance metric for emulation of rotational inertia



System norm:

amplification of

disturbances: impulse (fault), step (loss of unit), white noise (renewables)

performance outputs: integral, peak, ROCOF, restoration time, ...

19 / 35

Coherency performance metric & \mathcal{H}_2 norm

Energy expended by the system to return to synchronous operation:

$$\int_0^\infty \sum\nolimits_{\{i,j\}\in\mathcal{E}} a_{ij} (\theta_i(t) - \theta_j(t))^2 + \sum\nolimits_{i=1}^n s_i \, \omega_i^2(t) \, dt$$

 \mathcal{H}_2 norm interpretation:

1 associated **performance output**:

$$y = \begin{bmatrix} Q_1^{1/2} & 0 \\ 0 & Q_2^{1/2} \end{bmatrix} \begin{bmatrix} \theta \\ \omega \end{bmatrix}$$

- **2** impulses (faults) \longrightarrow output energy $\int_0^\infty y(t)^{\mathsf{T}} y(t) dt$
- **3 white noise** (renewables) \longrightarrow output variance $\lim_{t\to\infty} \mathbb{E}\left(y(t)^{\mathsf{T}}y(t)\right)$

Algebraic characterization of the \mathcal{H}_2 norm

Lemma: via observability Gramian

$$||G||_2^2 = \text{Trace}(B^{\mathsf{T}}PB)$$

where P is the observability Gramian $P = \int_0^\infty e^{A^T t} C^T C e^{At} dt$

- ▶ P solves a Lyapunov equation: $PA + A^TP + Q = 0$
- lacktriangledown A has a zero eigenvalue ightarrow restricts choice of Q

$$y = \begin{bmatrix} Q_1^{1/2} & 0 \\ 0 & Q_2^{1/2} \end{bmatrix} \begin{bmatrix} \theta \\ \omega \end{bmatrix} \qquad Q_1^{1/2} \mathbb{1} = 0$$

▶ P is unique for $P \begin{bmatrix} 1 & 0 \end{bmatrix} = \begin{bmatrix} 0 & 0 \end{bmatrix}$

Problem formulation

$$\begin{array}{ll} \underset{P,\,m_i}{\text{minimize}} & \|G\|_2^2 = \text{Trace}(B^\mathsf{T}PB) & \rightarrow \text{performance metric} \\ \text{subject to} & \sum_{i=1}^n m_i \leq m_{\text{bdg}} & \rightarrow \text{budget constraint} \\ & \underline{m_i} \leq m_i \leq \overline{m_i} \,, \quad i \in \{1,\dots,n\} \to \text{capacity constraint} \\ & PA + A^\mathsf{T}P + Q = 0 & \rightarrow \text{observability Gramian} \\ & P \left[\mathbb{1} \ \mathbb{0}\right] = \left[\mathbb{0} \ \mathbb{0}\right] & \rightarrow \text{uniqueness} \end{array}$$

Insights

- $oldsymbol{0}$ m appears as m^{-1} in system matrices A, B
- 2 product of B(m) & P in the objective
- 3 product of A(m) & P in the constraint

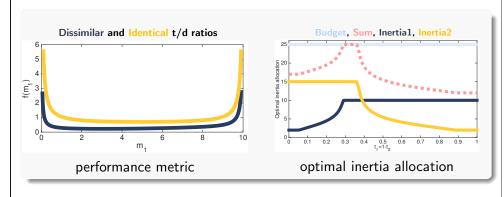
⇒ large-scale & non-convex

22 / 35

Building the intuition: results for two-area networks

Fundamental learnings

- explicit closed-form solution is rational function
- 2 sufficiently uniform $(t/d)_i \rightarrow \text{strongly convex } \& \text{ fairly flat cost}$
- 3 non trivial in the presence of capacity constraints



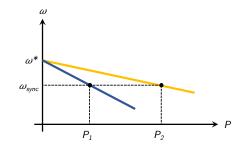
23 / 35

Closed-form results for cost of primary control

$P/\dot{\theta}$ primary droop control

$$(\omega_i - \omega^*) \propto (P_i^* - P_i(\theta))$$

$$\updownarrow$$
 $D_i \dot{\theta}_i = P_i^* - P_i(\theta)$



Primary control effort ightarrow accounted for by integral quadratic cost

$$\int_0^\infty \dot{\theta}(t)^\mathsf{T} D \,\dot{\theta}(t) \,dt$$

which is the \mathcal{H}_2 performance for the penalties $Q_1^{1/2}=0$ and $Q_2^{1/2}=D$

Primary Control ... cont'd

Theorem: the primary control effort optimization reads equivalently as

minimize
$$\sum_{i=1}^{n} \frac{t_i}{m_i}$$
 subject to
$$\sum_{i=1}^{n} m_i \leq m_{\text{bdg}}$$

$$\underline{m_i} \leq m_i \leq \overline{m_i}, \quad i \in \{1, \dots, n\}$$

Key take-aways:

- optimal solution independent of network topology
- ▶ allocation $\propto \sqrt{t_i}$ or $m_i = \min\{m_{\text{bdg}}, \overline{m_i}\}$

Location & strength of disturbance are crucial solution ingredients

numerical method for the general case

Taylor & power series expansions

Key idea: expand the performance metric as a power series in m

$$\|G\|_2^2 = \operatorname{Trace}(B(m)^T P(m)B(m))$$

Motivation: scalar series expansion at m_i in direction μ_i :

$$rac{1}{(m_i + oldsymbol{\delta}\mu_i)} = rac{1}{m_i} - rac{oldsymbol{\delta}\mu_i}{m_i^2} + \mathcal{O}(oldsymbol{\delta}^2)$$

Expand system matrices as **Taylor series** in direction μ :

$$\mathbf{A}(m+\delta\mu) = \mathbf{A}^{(0)}_{(m,\mu)} + \mathbf{A}^{(1)}_{(m,\mu)}\delta + \mathcal{O}(\delta^2)$$

$$\mathsf{B}(m+\delta\mu) = \mathsf{B}^{(0)}_{(m,\mu)} + \mathsf{B}^{(1)}_{(m,\mu)}\delta + \mathcal{O}(\delta^2)$$

Expand the observability Gramian as a **power series** in direction μ :

$$\mathbf{P}(m+\delta\mu) = \mathbf{P}^{(0)}_{(m,\mu)} + \mathbf{P}^{(1)}_{(\mathbf{m},\mu)} \delta + \mathcal{O}(\delta^2)$$

26 / 35

Explicit gradient computation

Expansion of system matrices & Gramian ⇒ match coefficients . . .

Formula for gradient at m in direction μ

1 nominal Lyapunov equation for $\mathcal{O}(\delta^0)$:

$$P^{(0)} = Lyap(A^{(0)}, Q)$$

2 perturbed Lyapunov equation for $\mathcal{O}(\delta^1)$ terms:

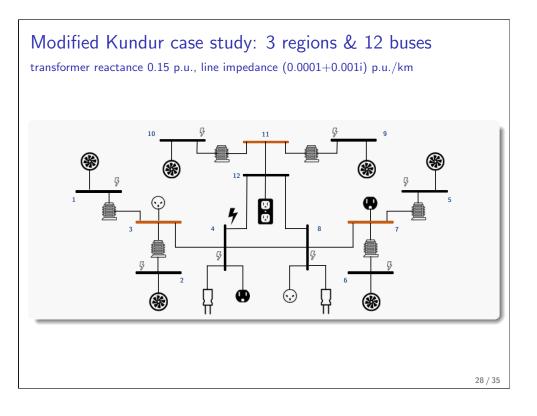
$$P^{(1)} = Lvap(A^{(0)}, P^{(0)}A^{(1)} + A^{(1)}^TP^{(0)})$$

3 expand objective in direction μ :

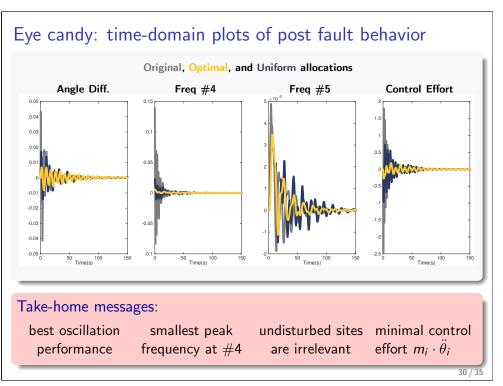
$$\|G\|_2^2 = \operatorname{Trace}(B(m)^{\mathsf{T}} \mathbf{P}(m) B(m)) = \operatorname{Trace}((\ldots) + \delta(\ldots)) + \mathcal{O}(\delta^2)$$

- **9** gradient: Trace $(2 * B^{(1)^T} P^{(0)} B^{(0)} + B^{(0)^T} P^{(1)} B^{(0)})$
- ⇒ use favorite method for reduced optimization problem

results



Heuristics outperformed by \mathcal{H}_2 - optimal allocation Original, Optimal, and Capacity allocations Cost Scenario: disturbance at #4 ▶ locally optimal solution outperforms heuristic max/uniform allocation ightharpoonup optimal allocation pproxallocation subject to capacity constraints matches disturbance Original, Optimal, and Uniform allocations inertia emulation at all undisturbed nodes is actually detrimental ⇒ **location** of disturbance & inertia emulation matters allocation subject to the budget constraint





Conclusions on virtual inertia emulation

Where to do it?

- $oldsymbol{0}$ \mathcal{H}_2 -optimal (non-convex) allocation
- closed-form results for cost of primary control
- numerical approach via gradient computation

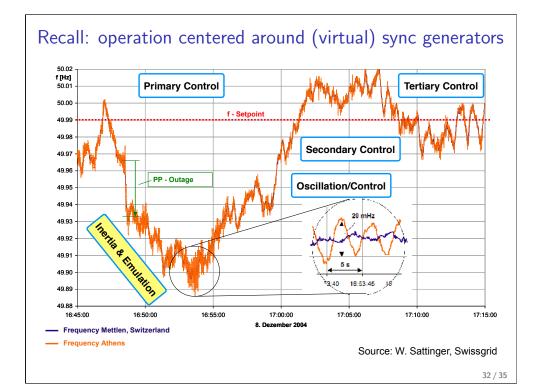
How to do it?

- down-sides of naive inertia emulation
- 2 novel machine matching control

What else to do? Inertia emulation is ...

- decentralized, plug-and-play (passive), grid-friendly, user-friendly, . . .
- suboptimal, wasteful in control effort, & need for new actuators

31 / 35



A control perspective of power system operation

Conventional strategy: emulate generator physics & control

$$M\dot{\omega}(t) = P_{\text{mech}} - D\omega(t) - \int_{0}^{t} \omega(\tau) d\tau - P_{\text{elec}}$$

(virtual) inertia tertiary control primary control secondary control

Essentially all PID + setpoint control (simple, robust, & scalable)

$$\underbrace{M\dot{\omega}(t)}_{\text{D}} = \underbrace{P}_{\text{set-point}} - \underbrace{D\omega(t)}_{\text{P}} - \underbrace{\int_{0}^{t} \omega(\tau) \, d\tau}_{\text{I}} - P_{\text{elec}}$$

Control engineers should be able to do better ...

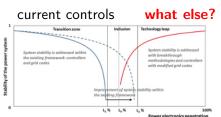


by TSOs, device manufacturers, academia, etc.

Massive InteGRATion of power Electronic devices



"The question that has to be examined is: how much power electronics can the grid cope with?" (European Commission)



all options are on the table and keep us busy ...

34 / 35

Acknowledgements



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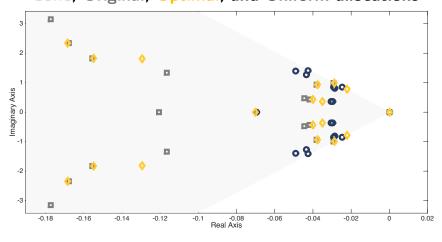


35 / 35

appendix

Spectral perspective on different inertia allocations

Cone, Original, Optimal, and Uniform allocations



• $\mathbf{m} = \mathbf{m} \rightarrow \text{best damping asymptote } \& \text{ best damping ratio}$

• Spectrum holds only partial information!!

The planning problem

sparse allocation of limited resources

 ℓ_1 -regularized inertia allocation (promoting a sparse solution):

minimize
$$\mathbf{J}_{\gamma}(\mathbf{m},\mathbf{P}) = \|G\|_2^2 + \frac{\mathbf{m}}{\|\mathbf{m} - \underline{\mathbf{m}}\|_1}$$
 subject to
$$\sum_{i=1}^n m_i \le m_{\mathrm{bdg}}$$

$$\underline{m_i} \le m_i \le \overline{m_i} \quad i \in \{1,\dots,n\}$$

$$PA + A^{\mathsf{T}}P + Q = 0$$

$$P[\mathbb{1} \ \mathbb{0}] = [\mathbb{0} \ \mathbb{0}]$$

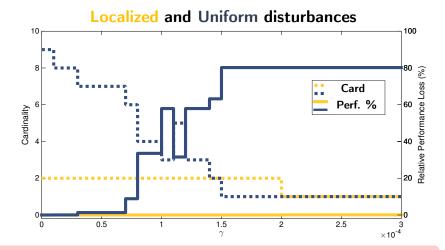
where $\gamma \geq 0$ trades off sparsity penalty and the original objective

Highlights:

- regularization term is linear & differentiable
- 2 gradient computation algorithm can be used with some tweaking

Relative performance loss (%) as a function of γ

0% o optimal allocation, 100% o no additional allocation



- **1** uniform disturbance $\Rightarrow \exists \gamma$ **1.3%** loss \equiv **(9** \rightarrow **7)**
- **2** localized disturbance \Rightarrow (2 \rightarrow 1) without affecting performance

Uniform disturbance to damping ratio

power sharing o $extbf{d} \propto P^*$, assuming $extbf{t} \propto$ source rating P^*

Theorem: for $t_i/d_i=t_j/d_j$ the allocation problem reads equivalently as

$$\begin{array}{ll} \text{minimize} & \sum_{i=1}^n \frac{s_i}{m_i} \\ \text{subject to} & \sum_{i=1}^n m_i \leq m_{\text{bdg}} \\ & \underline{m_i} \leq m_i \leq \overline{m_i}, \ i \in \{1,\dots,n\} \end{array}$$

Key takeaways:

- optimal solution independent of network topology
- allocation $\propto \sqrt{s_i}$ or $m_i = \min\{m_{\text{bdg}}, \overline{m_i}\}$

What if freq. penalty \propto inertia? \rightarrow norm independent of inertia

Taylor & power series expansions

Key idea: expand the performance metric as a power series in m

$$||G||_2^2 = \operatorname{Trace}(B(m)^{\mathsf{T}} \mathbf{P}(m) B(m))$$

Motivation: scalar series expansion at m_i in direction μ_i :

$$rac{1}{\left(m_i+oldsymbol{\delta}\mu_i
ight)}=rac{1}{m_i}-rac{oldsymbol{\delta}\mu_i}{m_i^2}+\mathcal{O}(oldsymbol{\delta}^2)$$

Expand system matrices in direction μ , where $\Phi = \operatorname{diag}(\mu)$:

$$\mathbf{A}_{(\mathbf{m},\mu)}^{(0)} = \begin{bmatrix} 0 & I \\ -M^{-1}L & -M^{-1}D \end{bmatrix}, \ \mathbf{A}_{(\mathbf{m},\mu)}^{(1)} = \begin{bmatrix} 0 & 0 \\ \Phi M^{-2}L & \Phi M^{-2}D \end{bmatrix}$$
$$\mathbf{B}_{(\mathbf{m},\mu)}^{(0)} = \begin{bmatrix} 0 \\ M^{-1}T^{1/2} \end{bmatrix}, \ \mathbf{B}_{(\mathbf{m},\mu)}^{(1)} = \begin{bmatrix} 0 \\ -\Phi M^{-2}T^{1/2} \end{bmatrix}$$

Taylor & power series expansions cont'd

Expand the observability Gramian as a power series in direction μ

$$\mathbf{P}(m) = \mathbf{P}(m + \delta\mu) = \mathbf{P}_{(m,\mu)}^{(0)} + \mathbf{P}_{(m,\mu)}^{(1)} \delta + \mathcal{O}(\delta^2)$$

Formula for gradient in direction μ

- **1** nominal Lyapunov equation for $\mathcal{O}(\delta^0)$: $\mathbf{P^{(0)}} = \mathbf{Lyap}(\mathbf{A^{(0)}}, \mathbf{Q})$
- 2 perturbed Lyapunov equation for $\mathcal{O}(\delta^1)$ terms:

$$\mathbf{P^{(1)}} = \mathsf{Lyap}(\mathbf{A^{(0)}}, \mathbf{P^{(0)}}\mathbf{A^{(1)}} + \mathbf{A^{(1)}}^\mathsf{T}\mathbf{P^{(0)}})$$

3 expand objective in direction μ :

$$||G||_2^2 = \operatorname{Trace}(B(m)^{\mathsf{T}} \mathbf{P}(m)B(m)) = \operatorname{Trace}((\ldots) + \delta(\ldots)) + \mathcal{O}(\delta^2)$$

4 gradient: $Trace(2 * B^{(1)^T}P^{(0)}B^{(0)} + B^{(0)^T}P^{(1)}B^{(0)})$

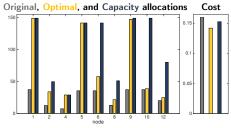
Gradient computation

Algorithm: Gradient computation & perturbation analysis

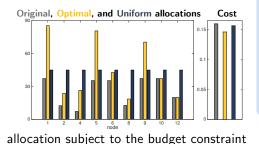
Input \rightarrow current values of the decision variables $\mathbf{m_i}$ Output \rightarrow numerically evaluated gradient ∇f of the cost function

- Evaluate the system matrices $A^{(0)}$, $B^{(0)}$ based on current inertia
- 2 Solve for $P^{(0)} = Lyap(A^{(0)}, Q)$ using a Lyapunov routine
- **3** For each node- obtain the perturbed system matrices $A^{(1)}$, $B^{(1)}$
- **4** Compute $P^{(1)} = Lyap(A^{(0)}, P^{(0)}A^{(1)} + A^{(1)^T}P^{(0)})$
- $\textbf{ G} \mathsf{ Gradient} \Rightarrow \textbf{Trace} \big(2 * \textbf{B}^{(1)}^\mathsf{T} \textbf{P}^{(0)} \textbf{B}^{(0)} + \textbf{B}^{(0)}^\mathsf{T} \textbf{P}^{(1)} \textbf{B}^{(0)} \big)$

Heuristics outperformed also for uniform disturbance



allocation subject to capacity constraints



Scenario: uniform disturbance

Heuristics for placement:

- max allocation in case of capacity constraints
- uniform allocation in case of budget constraint

Results & insights:

- locally optimal solution outperforms heuristics
- ② optimal solution ≠ max inertia at each bus

