



Advanced grid-forming control for low-inertia systems

Florian Dörfler

ETH Zürich

Emerging Topics in Control of Power Systems

Acknowledgements



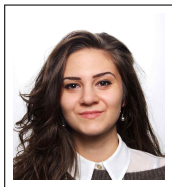
Marcello Colombino



Ali Tayyebi-Khameneh



Dominic Groß



Irina Subotic



ETH Foundation
Zürich



FONDS NATIONAL SUISSE
SCHWEIZERISCHER NATIONALFONDS
FONDO NAZIONALE SVIZZERO
SWISS NATIONAL SCIENCE FOUNDATION

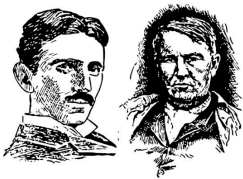


Eidgenössische Technische Hochschule Zürich
Swiss Federal Institute of Technology Zürich



Further: T. Jouini, C. Arghir, A. Anta, B. Johnson, M. Sinha, & S. Dhople

Replacing the power system foundation



fuel & synchronous machines

- not sustainable
- + central & dispatchable generation
- + large rotational inertia as buffer
- + self-synchronize through the grid
- + resilient voltage / frequency control
- slow actuation & control

renewables & power electronics

- + sustainable
- distributed & variable generation
- almost no energy storage
- no inherent self-synchronization
- fragile voltage / frequency control
- + fast / flexible / modular control

Critically re-visit modeling / analysis / control

Foundations and Challenges of Low-Inertia Systems

(Invited Paper)

Federico Milano
University College Dublin, Ireland
email: federico.milano@ucd.ie

Florian Dörfler and Gabriela Hug
ETH Zürich, Switzerland
emails: dorfler@ethz.ch,
ghug@ethz.ch

David J. Hill* and Gregor Verbič
University of Sydney, Australia
* also University of Hong Kong
emails: dhill@eee.hku.hk,
gregor.verbic@sydney.edu.au

The later sections contain many suggestions for further work, which can be summarized as follows:

- **New models** are needed which balance the need to include key features without burdening the model (whether for analytical or computational work) with uneven and excessive detail;
- **New stability theory** which properly reflects the new devices and time-scales associated with CIG, new loads and use of storage;
- Further **computational work** to achieve sensitivity guidelines including data-based approaches;
- **New control methodologies**, e.g. new controller to mitigate the high rate of change of frequency in low inertia systems;
- A power converter is a fully actuated, modular, and very fast control system, which are nearly antipodal characteristics to those of a synchronous machine. Thus, **one should critically reflect the control** of a converter as a virtual synchronous machine; and
- The lack of inertia in a power system does not need to (and **cannot**) be fixed by simply “adding inertia back” in the systems.

key unresolved challenge: resilient control of grid-forming power converters

→ industry & academia willing to explore **green-field approach** (see MIGRATE)

Outline

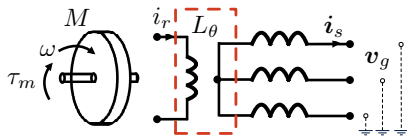
Introduction: Low-Inertia Power Systems

Problem Setup: Modeling and Specifications

State of the Art Grid-Forming Control

Comparison & Discussion

Modeling: synchronous generator



$$\frac{d\theta}{dt} = \omega$$

$$M \frac{d\omega}{dt} = -D\omega + \tau_m + L_m i_r \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix}^\top \mathbf{i}_s$$

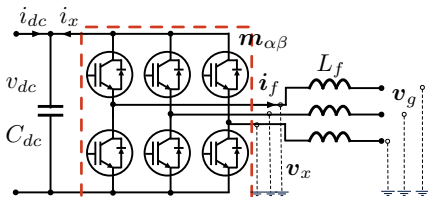
$$L_s \frac{d\mathbf{i}_s}{dt} = -R_s \mathbf{i}_s + \mathbf{v}_g - L_m i_r \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix} \omega$$

1. **energy supply** τ_m from governor
2. mechanical (θ, ω) **swing dynamics** of rotor (flywheel) with inertia M
3. i_s **stator flux dynamics** (rotor/damper flux dynamics neglected)
4. **electro-mechanical energy conversion** through rotating magnetic field with inductance matrix

$$L_\theta = \begin{bmatrix} L_s & 0 & L_m \cos \theta \\ 0 & L_s & L_m \sin \theta \\ L_m \cos \theta & L_m \sin \theta & L_r \end{bmatrix}$$

Modeling: voltage source converter

1. **energy supply** i_{dc} from upstream DC boost converter
2. **DC link dynamics** v_{dc} with capacitance C_{dc}
3. **i_f AC filter dynamics** (sometimes also LC or LCL)



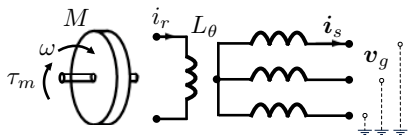
4. **power electronics modulation**

$$i_x = -\mathbf{m}^\top \mathbf{i}_f \quad \text{and} \quad \mathbf{v}_x = \mathbf{m} v_{dc},$$

with averaged & normalized duty cycle ratios $\mathbf{m} \in [-\frac{1}{2}, \frac{1}{2}] \times [-\frac{1}{2}, \frac{1}{2}]$

$$C_{dc} \frac{dv_{dc}}{dt} = -G_{dc} v_{dc} + i_{dc} + \mathbf{m}^\top \mathbf{i}_f$$
$$L_f \frac{d\mathbf{i}_f}{dt} = -R_f \mathbf{i}_f + \mathbf{v}_g - \mathbf{m} v_{dc}$$

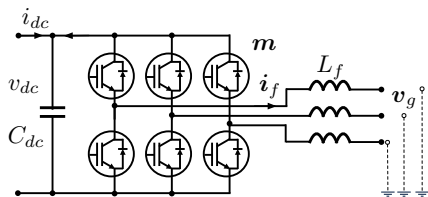
Comparison: conversion mechanisms



$$\frac{d\theta}{dt} = \omega$$

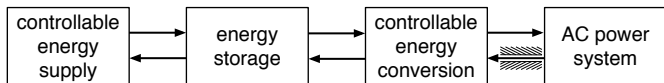
$$M \frac{d\omega}{dt} = -D\omega + \tau_m + L_m i_r \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix}^\top \mathbf{i}_s$$

$$L_s \frac{d\mathbf{i}_s}{dt} = -R_s \mathbf{i}_s + \mathbf{v}_g - L_m i_r \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix} \omega$$



$$C_{dc} \frac{dv_{dc}}{dt} = -G_{dc} v_{dc} + i_{dc} + \mathbf{m}^\top \mathbf{i}_f$$

$$L_f \frac{d\mathbf{i}_f}{dt} = -R_f \mathbf{i}_f + \mathbf{v}_g - \mathbf{m} v_{dc}$$



τ_m (slow)

vs.

i_{dc} (fast)

M (large)

vs.

C_{dc} (small)

L_θ (physical)

vs.

m (control)

resilient

vs.

fragile

(over-currents)

physical & robust
vs.

controlled & agile

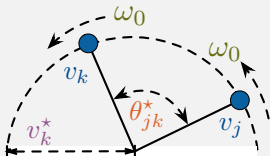
**signal / energy
transformer**

Objectives for grid-forming converter control ($\alpha\beta$ frame)

stationary control objectives

- ▶ synchronous frequency

$$\frac{d}{dt} v_k = \begin{bmatrix} 0 & -\omega_0 \\ \omega_0 & 0 \end{bmatrix} v_k$$



- ▶ voltage amplitude $\|v_k\| = v_k^*$

- ▶ active & reactive power injections

$$v_k^\top i_{f,k} = p_k^* \quad , \quad v_k^\top \begin{bmatrix} 0 & -1 \\ +1 & 0 \end{bmatrix} i_{f,k} = q_k^*$$

unique
 \iff relative voltage angles
conversion

$$v_k = \begin{bmatrix} \cos(\theta_{jk}^*) & -\sin(\theta_{jk}^*) \\ \sin(\theta_{jk}^*) & \cos(\theta_{jk}^*) \end{bmatrix} v_j$$

dynamic control objectives

- ▶ droop at perturbed operation: $\omega - \omega_0 = k \cdot (p - p^*)$ with specified power/frequency sensitivity $k = \frac{\partial p}{\partial \omega}$ droop (similar for $\|v\|$ and q)
- ▶ disturbance (fault) rejection: passively via physics (inertia) or via control
- ▶ grid-forming: intrinsic synchronization rather than tracking of exogenous ω_0

Naive baseline solution: emulation of virtual inertia

IRELAND

Hybrid storage system looks to Ireland's services market

22 November 2016 by Sara Verbruggen · Be the first to comment

IRELAND: The pilot of a 576kW grid storage system using flywheels and batteries by Dublin-based Schwungrad Energie is the first of its technology's deployment in Ireland's ancillary services market.



Flywheel storage provide

Can Synthetic Inertia from Wind Power Stabilize Grids?

By Peter Fairley
Posted 7 Nov 2016 | 21:00 GMT



Pure-play battery or hybrid grid energy storage?

Oct 11, 2016 12:54 PM BST

Share



Improvement of Transient Response in Microgrids Using Virtual Inertia

Nimish Soni, Student Member, IEEE, Suryanarayana Doolla, Member, IEEE, and Mukul C. Chandorkar, Member, IEEE

Virtual synchronous generators: A survey and new perspectives

Hassan Bevrani^{a,b,v}, Toshifumi Ise^b, Yushi Miura^b

^aDept. of Electrical and Computer Eng., University of Kurdistan, PO Box 416, Sanandaj, Iran

^bDept. of Electrical, Electronic and Information Eng., Osaka University, Osaka, Japan

IEEE TRANSACTIONS ON POWER SYSTEMS, VOL. 28, NO. 2, MAY 2013

Inertia Emulation Control Strategy for VSC-HVDC Transmission Systems

Jiebei Zhu, Campbell D. Booth, Grain P. Adam, Andrew J. Roscoe, and Chris G. Bright

Implementing Virtual Inertia in DFIG-Based Wind Power Generation

Mahmadsreza Fakhari Moghaddam Arani, Student Member, IEEE, and Ehab F. El-Saadany, Senior Member, IEEE

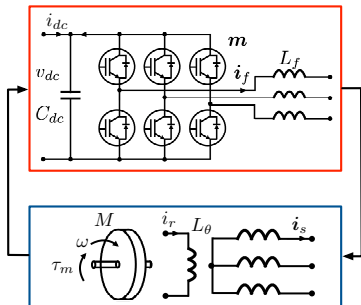
Dynamic Frequency Control Support: a Virtual Inertia Provided by Distributed Energy Storage to Isolated Power Systems

Gauthier Delille, Member, IEEE, Bruno François, Senior Member, IEEE, and Gilles Malarange

Grid Tied Converter with Virtual Kinetic Storage

M.P.N van Wessenbeeck¹, S.W.H. de Haan¹, Senior member, IEEE, P. Varella² and K. Visscher³,

Virtual synchronous machine \equiv flywheel emulation



- **reference model** for converter voltage loop: detailed model of synchronous generator + controls (of order 3, ..., 12)

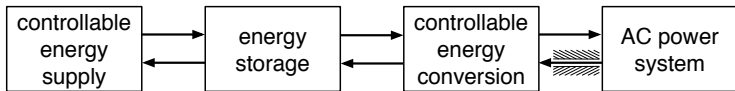
→ most commonly **accepted solution** in **industry** (¿ backward compatibility ?)

- robust **implementation needs tricks**: low-pass filters for dissipation, virtual impedances for saturation, limiters, ...

→ performs well in small-signal regime but **performs very poorly post-fault**

→ **poor fit**: converter \neq flywheel
very different actuation & energy storage

→ **over-parametrized** & ignores **limits**



slow vs. fast

large vs. small

physics vs. control

resilient vs. fragile

Droop as simplest reference model

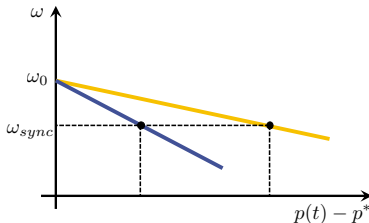
[Chandorkar, Divan, Adapa, '93]

- ▶ **frequency control** by mimicking $p - \omega$ droop property of synchronous machine:

$$\omega - \omega_0 \propto p - p^*$$

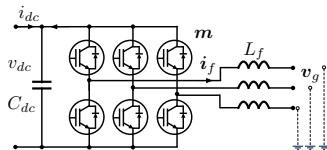
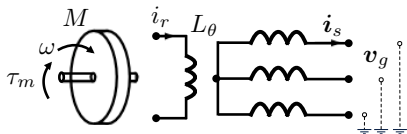
- ▶ **voltage control** via $q - \|v\|$ droop control:

$$\frac{d}{dt} \|v\| = -c_1(\|v\| - v^*) - c_2(q - q^*)$$



- **reference** are generator controls
- direct control of (p, ω) and $(q, \|v\|)$ **assuming they are independent** (approx. true only near steady state)
- requires **tricks in implementation**: similar to virtual synchronous machine
- **good small-signal but poor large signal behavior** (rather narrow region of attraction)
- main reason for poor performance: **two linear SISO loops for MIMO nonlinear system** (SISO & linear only near steady state)

Duality & matching of synchronous machines [Arghir & Dörfler, '19]



$$\frac{d\theta}{dt} = \omega$$

$$M \frac{d\omega}{dt} = -D\omega + \tau_m + L_m i_r \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix}^\top \mathbf{i}_s$$

$$L_s \frac{d\mathbf{i}_s}{dt} = -R_s \mathbf{i}_s + \mathbf{v}_g - L_m i_r \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix} \omega$$

$$\frac{d\theta}{dt} = \eta \cdot v_{dc}$$

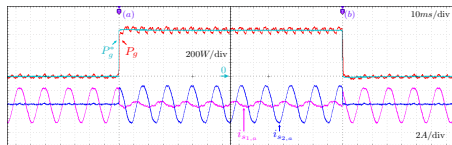
$$C_{dc} \frac{dv_{dc}}{dt} = -G_{dc} v_{dc} + i_{dc} + m_{\text{ampl}} \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix}^\top \mathbf{i}_f$$

$$L_f \frac{d\mathbf{i}_f}{dt} = -R_f \mathbf{i}_f + \mathbf{v}_g - m_{\text{ampl}} \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix} v_{dc}$$

1. modulation in polar coordinates:

$$m = m_{\text{ampl}} \begin{bmatrix} -\sin \theta \\ \cos \theta \end{bmatrix} \quad \& \quad \dot{\theta} = m_{\text{freq}}$$

2. **matching**: $m_{\text{freq}} = \eta v_{dc}$ with $\eta = \frac{\omega_{\text{ref}}}{v_{dc, \text{ref}}}$

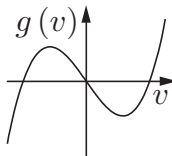
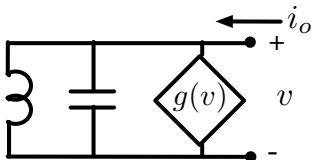


→ **duality**: $C_{dc} \sim M$ is equivalent inertia

theory & practice: **robust** duality $\omega \sim v_{dc}$ 11

Original Virtual Oscillator Control (VOC)

nonlinear & open limit cycle oscillator as reference model for converter voltage loop



- simplified model amenable to theoretic analysis

[J. Aracil & F. Gordillo, '02], [Torres, Hespanha, Moehlis, '11],
[Johnson, Dhople, Krein, '13], [Dhople, Johnson, Dörfler, '14]

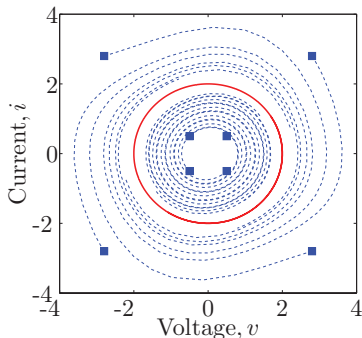
→ *almost global synchronization & local droop*

- in practice proven to be *robust mechanism* with performance superior to droop & others

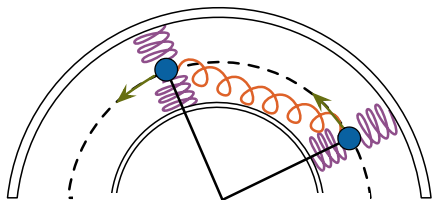
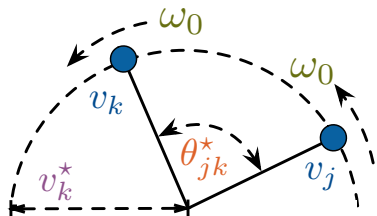
→ **problem**: cannot be controlled(?) to meet specifications on amplitude & power injections

→ *dispatchable* virtual oscillator control

[Colombino, Groß, Brouillon, & Dörfler, '17, '18, '19],
[Subotic, Gross, Colombino, & Dörfler, '19]



Colorful idea: closed-loop target dynamics



$$\frac{d}{dt} \mathbf{v}_k = \underbrace{\begin{bmatrix} 0 & -\omega_0 \\ \omega_0 & 0 \end{bmatrix} \mathbf{v}_k}_{\text{rotation at } \omega} + \underbrace{c_2 \cdot (\|\mathbf{v}_k\|^{*2} - \|\mathbf{v}_k\|^2) \mathbf{v}_k}_{\text{amplitude regulation to } v_k^*}$$

$$+ c_1 \cdot \underbrace{\sum_{j=1}^n w_{jk} \left(\mathbf{v}_j - \begin{bmatrix} \cos(\theta_{jk}^*) & -\sin(\theta_{jk}^*) \\ \sin(\theta_{jk}^*) & \cos(\theta_{jk}^*) \end{bmatrix} \mathbf{v}_k \right)}_{\text{synchronization to desired relative angles } \theta_{jk}^*}$$

Properties of virtual oscillator control

1. desired target dynamics can be realized via **fully decentralized control**

$$\frac{d}{dt} v_k = \underbrace{\begin{bmatrix} 0 & -\omega_0 \\ \omega_0 & 0 \end{bmatrix} v_k}_{\text{rotation at } \omega_0} + c_1 \cdot \underbrace{\left(\frac{1}{v_k^{*2}} \begin{bmatrix} q_k^* & p_k^* \\ -p_k^* & q_k^* \end{bmatrix} v_k - i_{f,k} \right)}_{\text{synchronization through grid current}} + c_2 \cdot \underbrace{(v_k^{*2} - \|v_k\|^2)}_{\text{local amplitude regulation}} v_k$$

2. connection to **droop control** seen in polar coordinates (though multivariable)

$$\frac{d}{dt} \theta_k = \omega_0 + c_1 \left(\frac{p_k^*}{v_k^{*2}} - \frac{p_k}{\|v_k\|^2} \right) \Big|_{\|v_k\| \approx 1} \approx \omega_0 + c_1 (p_k^* - p_k) \quad (p - \omega \text{ droop})$$

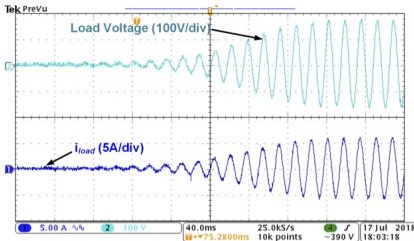
$$\frac{d}{dt} \|v_k\| \Big|_{\|v_k\| \approx 1} \approx c_1 (q_k^* - q_k) + c_2 (v_k^* - \|v_k\|) \quad (q - \|v\| \text{ droop})$$

3. **almost global asymptotic stability** with respect to pre-specified set-point if

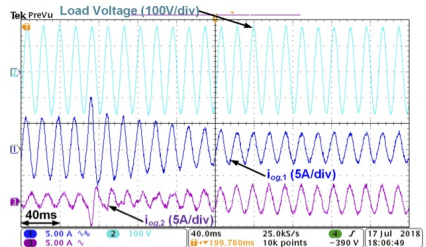
- ▶ **power transfer** “small” compared to **network connectivity**
- ▶ **amplitude control** “slower” than **synchronization control**

Experimental results

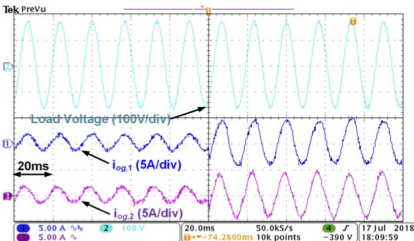
[Seo, Subotic, Johnson, Colombino, Groß, & Dörfler, '19]



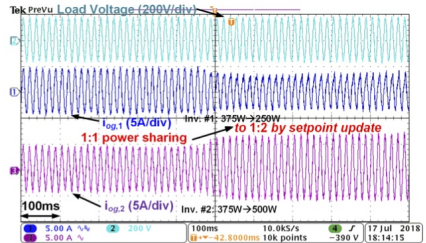
black start of inverter #1 under 500 W load
(making use of almost global stability)



connecting inverter #2 while inverter #1 is
regulating the grid under 500 W load

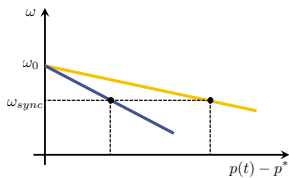


250 W to 750 W load transient with two
inverters active



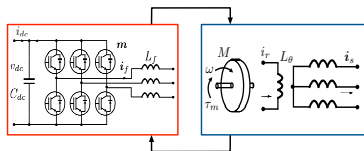
change of setpoint: p^* of inverter #2
updated from 250 W to 500 W

High-level comparison of grid-forming control



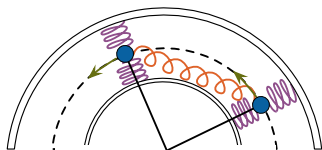
droop control

- + good performance near steady state
- relies on decoupling & small attraction basin



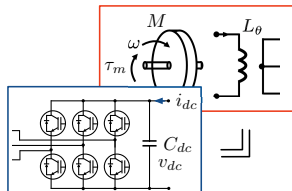
synchronous machine emulation

- + backward compatible in nominal case
- not resilient under large disturbances



virtual oscillator control

- + excellent large-signal behavior + local droop



matching control & duality

- + simple & robust

Detailed comparison(s) of control strategies

Comparison of Virtual Oscillator and Droop Controlled Isolated Three-Phase Microgrids

Zhan Shi ¹, Member, IEEE, Jiacheng Li ², Student Member, IEEE, Hendra I. Nurdin ³, Senior Member, IEEE, and John E. Fletcher ⁴, Senior Member, IEEE

Comparison of Virtual Oscillator and Droop Control

Brian Johnson, Miguel Rodriguez
Power Systems Engineering Center
National Renewable Energy Laboratory
Golden, CO 80401
Email: brian.johnson@nrel.gov, miguelr@gp@gmail.com

Mohit Sinha, Sairaj Dhole
Department of Electrical & Computer Engineering
University of Minnesota
Minneapolis, MN 55455
Email: {sinha952,sdhole1}@umn.edu

Similarities between Virtual Oscillator Controlled and Droop Controlled Three-Phase Inverters

Zhan Shi, Hendra I. Nurdin, John E. Fletcher, Jiacheng Li
School of Electrical Engineering and Telecommunications, UNSW Sydney, NSW, 2052, Australia
Email: zhan.shi@unsw.edu.au, h.nurdin@unsw.edu.au, john.fletcher@unsw.edu.au, jiacheng.li@unsw.edu.au

Comparative Transient Stability Assessment of Droop and Dispatchable Virtual Oscillator Controlled Grid-Connected Inverters

Hui Yu, Student Member, IEEE, M A Awal, Student Member, IEEE, Hao Tu, Student Member, IEEE, Iqbal Husain, Fellow, IEEE and Sanjan Lalkic, Senior Member, IEEE

Frequency Stability of Synchronous Machines and Grid-Forming Power Converters

Ali Tayyebi, Dominic Geof, Member, IEEE, Adolfo Anta, Friederich Kropog and Florian Dittler, Member, IEEE

GRID-FORMING CONVERTERS – INEVITABILITY, CONTROL STRATEGIES AND CHALLENGES IN FUTURE GRIDS APPLICATION

Ali TAYYEBI Florian DÖRFLER Friederich KUPZOG
AIT and ETH Zürich – Austria ETH Zürich – Switzerland Austrian Institute of Technology – Austria

Comparison of Droop Control and Virtual Oscillator Control Realized by Andronov-Hopf Dynamics

Minghui Lu¹, Victor Puub¹, Sairaj Dhole¹, Brian Johnson²

Transient response comparison of virtual oscillator controlled and droop controlled three-phase inverters under load changes

Zhan Shi¹, Jiacheng Li¹, Hendra I. Nurdin¹, John E. Fletcher¹
School of Electrical Engineering and Telecommunications, UNSW Sydney, NSW, 2052, Australia
E-mail: zhan.shi@unsw.edu.au

Simulation-based study of novel control strategies for inverters in low-inertia system: grid-forming and grid-following

Author: Alessandro Cristofalo

Mathias Melby

Comparison of virtual oscillator control and droop control in an inverter-based stand-alone microgrid

Grid-Forming Converters control based on DC voltage feedback

Yuan Guo¹, Hai-Peng Ren², Jie Li²

- ▶ **identical steady-state & similar small-signal behavior** (after tuning)
- ▶ **virtual synchronous machine** has poor transients (converter \neq flywheel)
- ▶ **VOC has best large-signal behavior**: stability, post-fault-response, ...
- ▶ **matching control** $\omega \sim v_{dc}$ **is most robust** though with slow AC dynamics
- ▶ ... comparison suggests **hybrid VOC + matching control** direction

Hybrid angle control = matching + oscillator control

hybrid angle control dynamics

$$\dot{\theta} = \omega_0 + \underbrace{c_1 \cdot (v_{dc} - v_{dc}^*)}_{\text{matching control term}} + \underbrace{c_2 \cdot \sin\left(\frac{\theta - \theta_{grid} - \theta^*}{2}\right)}_{\frac{1}{2} \text{ synchronizing oscillator term}}$$

a few selected *theoretical certificates*

- ▶ **almost global stability** for sufficiently large c_2/c_1
- ▶ **compatibility**: local droop behavior & stability preserved under dc source or ac grid dynamics
- ▶ active **current limitation** (pulling down modulation magnitude) with guaranteed closed-loop stability

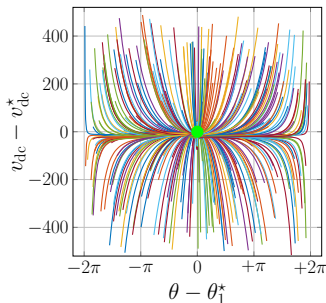
Hybrid Angle Control and Almost Global Stability of Grid-Forming Power Converters

Ali Tayyebi, Adolfo Anta, and Florian Dörfler

theory: **best grid-forming control** (!) → ongoing work: practice

implementation aspects

- ▶ **tuning gains**: c_1 & c_2 (robustness & performance)
- ▶ error signals ← **voltage & current measurements**
- ▶ θ either **voltage reference** or **modulation angle**



Exciting research bridging communities

